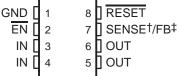
SLVS124F - JUNE 1995 - REVISED JANUARY 1999

- Available in 2.5-V, 3-V, 3.3-V, 4.85-V, and 5-V
   Fixed-Output and Adjustable Versions
- Integrated Precision Supply-Voltage Supervisor Monitoring Regulator Output Voltage
- Active-Low Reset Signal with 200-ms Pulse Width
- Very Low Dropout Voltage . . . Maximum of 35 mV at I<sub>O</sub> = 100 mA (TPS7350)
- Low Quiescent Current Independent of Load . . . 340 μA Typ
- Extremely Low Sleep-State Current,
   0.5 μA Max
- 2% Tolerance Over Full Range of Load, Line, and Temperature for Fixed-Output Versions§
- Output Current Range of 0 mA to 500 mA
- TSSOP Package Option Offers Reduced Component Height For Critical Applications

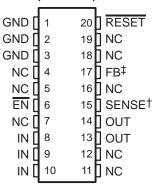
### description

The TPS73xx devices are members of a family of micropower low-dropout (LDO) voltage regulators.

# D OR P PACKAGE (TOP VIEW) D 1 8 RESE



### PW PACKAGE (TOP VIEW)



NC - No internal connection

† SENSE – Fixed voltage options only

(TPS7325, TPS7330, TPS7333, TPS7348, and TPS7350)

‡FB – Adjustable version only (TPS7301)

They are differentiated from the TPS71xx and TPS72xx LDOs by their integrated delayed microprocessor-reset function. If the precision delayed reset is not required, the TPS71xx and TPS72xx should be considered. ¶

#### **AVAILABLE OPTIONS**

ТЛ	OUTP	JT VOL	TAGE		/E-GOING OLD VOLT		Р	ACKAGED DEVI	CES	CHIP FORM
	MIN	TYP	MAX	MIN	TYP	MAX	SMALL OUTLINE (D)	PLASTIC DIP (P)	TSSOP (PW)	(Y)
	4.9	5	5.1	4.55	4.65	4.75	TPS7350QD	TPS7350QP	TPS7350QPW	TPS7350Y
	4.75	4.85	4.95	4.5	4.6	4.7	TPS7348QD	TPS7348QP	TPS7348QPW	TPS7348Y
-40°C to	3.23	3.3	3.37	2.868	2.934	3	TPS7333QD	TPS7333QP	TPS7333QPW	TPS7333Y
125°C	2.94	3	3.06	2.58	2.64	2.7	TPS7330QD	TPS7330QP	TPS7330QPW	TPS7330Y
	2.425	2.5	2.575	2.23	2.32	2.39	TPS7325QD	TPS7325QP	TPS7325QPW	TPS7325Y
		djustable V to 9.7		1.101	1.123	1.145	TPS7301QD	TPS7301QP	TPS7301QPW	TPS7301Y

The D and PW packages are available taped and reeled. Add an R suffix to device type (e.g., TPS7350QDR). The TPS7301Q is programmable using an external resistor divider (see application information). The chip form is tested at 25°C.

The TPS71xx and the TPS72xx are 500-mA and 250-mA output regulators respectively, offering performance similar to that of the TPS73xx but without the delayed-reset function. The TPS72xx devices are further differentiated by availability in 8-pin thin-shrink small-outline packages (TSSOP) for applications requiring minimum package size.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



<sup>§</sup> The TPS7325 has a tolerance of  $\pm 3\%$  over the full temperature range.

SLVS124F - JUNE 1995 - REVISED JANUARY 1999

### description (continued)

The RESET output of the TPS73xx initiates a reset in microcomputer and microprocessor systems in the event of an undervoltage condition. An internal comparator in the TPS73xx monitors the output voltage of the regulator to detect an undervoltage condition on the regulated output voltage.

If that occurs, the  $\overline{\text{RESET}}$  output (open-drain NMOS) turns on, taking the  $\overline{\text{RESET}}$  signal low.  $\overline{\text{RESET}}$  stays low for the duration of the undervoltage condition. Once the undervoltage condition ceases, a 200-ms (typ) time-out begins. At the completion of the 200-ms delay,  $\overline{\text{RESET}}$  goes high.

An order of magnitude reduction in dropout voltage and quiescent current over conventional LDO performance is achieved by replacing the typical pnp pass transistor with a PMOS device.

Because the PMOS device behaves as a low-value resistor, the dropout voltage is very low (maximum of 35 mV at an output current of 100 mA for the TPS7350) and is directly proportional to the output current (see Figure 1). Additionally, since the PMOS pass element is a voltage-driven device, the quiescent current is low and remains constant, independent of output loading (typically 340 µA over the full range of output current, 0 mA to 500 mA). These two key specifications yield a significant improvement in operating life for battery-powered systems.

The LDO family also features a sleep mode; applying a logic high signal to  $\overline{EN}$  (enable) shuts down the regulator, reducing the quiescent current to 0.5  $\mu$ A maximum at  $T_J = 25^{\circ}$ C.

The TPS73xx is offered in 2.5-V, 3-V, 3.3-V, 4.85-V, and 5-V fixed-voltage versions and in an adjustable version (programmable over the range of 1.2 V to 9.75 V). Output voltage tolerance is specified as a maximum of 2% over line, load, and temperature ranges (3% for the 2.5 V and the adjustable version). The TPS73xx family is available in PDIP (8 pin), SO (8 pin) and TSSOP (20 pin) packages. The TSSOP has a maximum height of 1.2 mm.

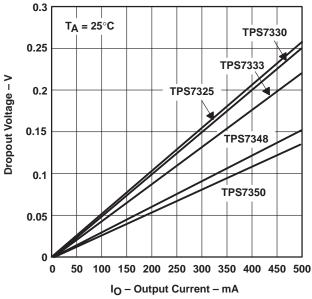
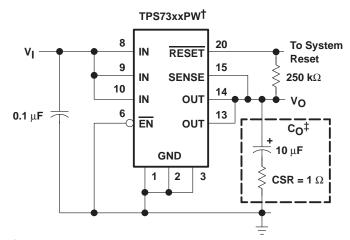


Figure 1. Dropout Voltage Versus Output Current



† TPS7325, TPS7330, TPS7333, TPS7348, TPS7350 (fixed-voltage options)

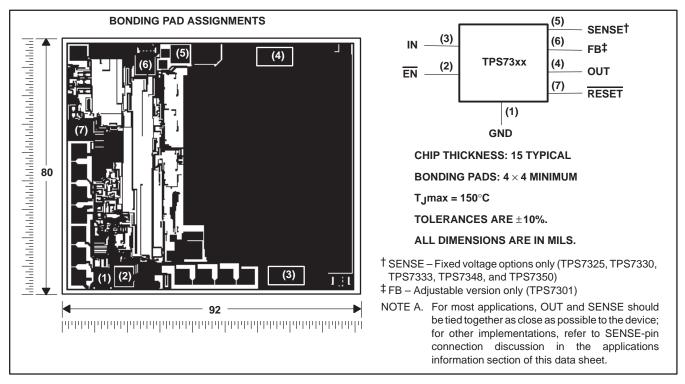
Figure 2. Typical Application Configuration



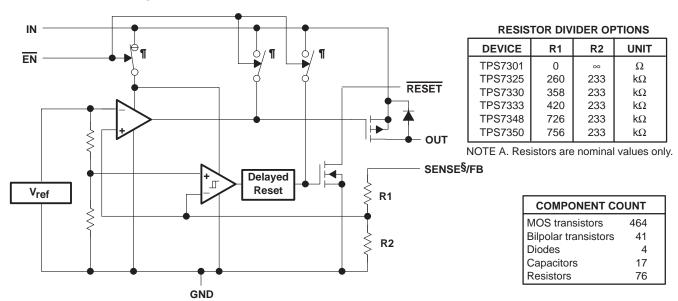
<sup>‡</sup> Capacitor selection is nontrivial. See application information section for details.

### **TPS73xxY** chip information

These chips, when properly assembled, display characteristics similar to those of the TPS73xxQ. Thermal compression or ultrasonic bonding may be used on the doped aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



### functional block diagram



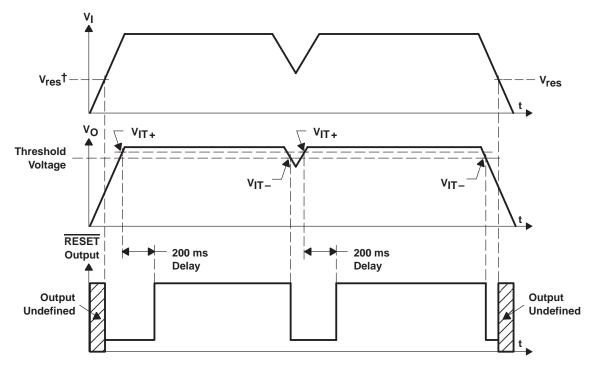
<sup>§</sup> For most applications, SENSE should be externally connected to OUT as close as possible to the device. For other implementations, refer to \_SENSE-pin connection discussion in applications information section.

<sup>¶</sup> Switch positions are shown with  $\overline{\mathsf{EN}}$  low (active).



SLVS124F – JUNE 1995 – REVISED JANUARY 1999

### timing diagram



<sup>†</sup> V<sub>res</sub> is the minimum input voltage for a valid RESET. The symbol V<sub>res</sub> is not currently listed within EIA or JEDEC standards for semiconductor symbology.

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)‡

Input voltage range§, V <sub>I</sub> , RESET, SENSE, EN	0.3 V to 11 V
Output current, I <sub>O</sub>	2 A
Continuous total power dissipation	See Dissipation Rating Tables 1 and 2
Operating virtual junction temperature range, T <sub>J</sub>	55°C to 150°C
Storage temperature range, T <sub>stg</sub>	65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

<sup>‡</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.



<sup>§</sup> All voltage values are with respect to network terminal ground.

SLVS124F - JUNE 1995 - REVISED JANUARY 1999

### DISSIPATION RATING TABLE 1 - FREE-AIR TEMPERATURE (SEE FIGURE 3)

PACKAGE	$T_{\mbox{$\mbox{$A$}}} \leq 25^{\circ}\mbox{$\mbox{$C$}}$ POWER RATING			T <sub>A</sub> = 125°C POWER RATING
D	725 mW	5.8 mW/°C	464 mW	145 mW
Р	1175 mW	9.4 mW/°C	752 mW	235 mW
PW†	700 mW	5.6 mW/°C	448 mW	140 mW

### DISSIPATION RATING TABLE 2 - CASE TEMPERATURE (SEE FIGURE 4)

PACKAGE	$T_C \le 25^{\circ}C$ POWER RATING	DERATING FACTOR ABOVE T <sub>C</sub> = 25°C	T <sub>C</sub> = 70°C POWER RATING	T <sub>C</sub> = 125°C POWER RATING
D	2188 mW	9.4 mW/°C	1765 mW	1248 mW
Р	2738 mW	21.9 mW/°C	1752 mW	548 mW
PWT	4025 mW	32.2 mW/°C	2576 mW	805 mW

<sup>†</sup> Refer to Thermal Information section for detailed power dissipation considerations when using the TSSOP package.

### MAXIMUM CONTINUOUS DISSIPATION

### VS FREE-AIR TEMPERATURE 1400 $\mathsf{P}_D-\mathsf{Maximum}$ Continuous Dissipation – mW 1200 1000 P Package $R_{\theta JA} = \bar{106}^{\circ}C/W$ 800 D Package 600 $R_{\theta JA} = 172^{\circ}C/W$ 400 PW Package $R_{\theta JA} = 178^{\circ} C/W$ 200 0 25 50 75 100 125 150 $T_A$ – Free-Air Temperature – $^{\circ}C$ Figure 3

### MAXIMUM CONTINUOUS DISSIPATION

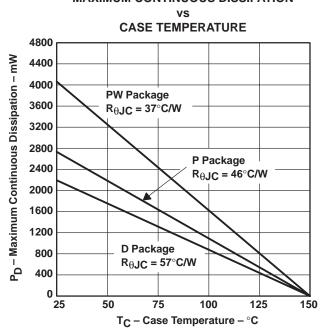


Figure 4

SLVS124F – JUNE 1995 – REVISED JANUARY 1999

### recommended operating conditions

		MIN	MAX	UNIT
	TPS7301Q	2.47	10	V
Input voltage, V <sub>I</sub> †	TPS7325Q	3.1	10	V
	TPS7330Q	3.5	10	V
	TPS7333Q	3.77	10	
	TPS7348Q	5.2	10	V
	TPS7350Q	5.33	10	
High-level input voltage at EN,	VIH	2		V
Low-level input voltage at EN,	/IL		0.5	V
Output current range, IO		0	500	mA
Operating virtual junction temp	erature range, TJ	-40	125	°C

<sup>†</sup> Minimum input voltage defined in the recommended operating conditions is the maximum specified output voltage plus dropout voltage, V<sub>DO</sub>, at the maximum specified load range. Since dropout voltage is a function of output current, the usable range can be extended for lighter loads. To calculate the minimum input voltage for the maximum load current used in a given application, use the following equation:

$$V_{I(min)} = V_{O(max)} + V_{DO(max load)}$$

Because the TPS7301 is programmable,  $r_{DS(on)}$  should be used to calculate  $V_{DO}$  before applying the above equation. The equation for calculating  $V_{DO}$  from  $r_{DS(on)}$  is given in Note 2 in the TPS7301 electrical characteristics table. The minimum value of 2.97 V is the absolute lower limit for the recommended input voltage range for the TPS7301.



SLVS124F - JUNE 1995 - REVISED JANUARY 1999

# electrical characteristics at I $_{O}$ = 10 mA, $\overline{\text{EN}}$ = 0 V, C $_{o}$ = 4.7 $\mu\text{F}$ (CSR $^{\ddagger}$ = 1 $\Omega$ ), SENSE/FB shorted to OUT (unless otherwise noted)

PARAMETER	TEST CO	NDITIONS§	TJ	MIN	TYP	MAX	UNIT	
Ground current (active mode)	<u>EN</u> ≤ 0.5 V,	$V_{I} = V_{O} + 1 V_{r}$	25°C		340	400		
Ground current (active mode)	$0 \text{ mA} \le I_{O} \le 500 \text{ m}$		-40°C to 125°C			550	μΑ	
Input current (standby mode)	ENL V	071/41/401/	25°C		0.01	0.5	μА	
Imput current (standby mode)	$EN = V_I$ ,	$2.7 \text{ V} \le \text{V}_{\text{I}} \le 10 \text{ V}$	-40°C to 125°C			2	μΑ	
Output current limit	V <sub>O</sub> = 0 V,	V <sub>I</sub> = 10 V	25°C		1.2	2	Α	
Output current innit	VO = 0 V,	V  = 10 V	-40°C to 125°C			2	_ ^	
Pass-element leakage current in standby	<u>-N</u> //	271/21/2401/	25°C		0.01	0.5	μА	
mode	$\overline{EN} = V_{I},$	2.7 V ≤ V <sub>I</sub> ≤ 10 V	-40°C to 125°C			1	μΑ	
DECET leake as surrent	None of an enetice	V et DECET 40 V	25°C		0.02	0.5	μА	
RESET leakage current	Normal operation,	V at RESET = 10 V	-40°C to 125°C			0.5	μΑ	
Output voltage temperature coefficient			-40°C to 125°C		61	75	ppm/°C	
Thermal shutdown junction temperature					165		°C	
CN la mia himb (atau dhu ma ada)	2.5 V ≤ V <sub>I</sub> ≤ 6 V		-40°C to 125°C	2			V	
EN logic high (standby mode)	$6 \text{ V} \leq \text{V}_{I} \leq 10 \text{ V}$		-40 C to 125 C	2.7			1 <sup>v</sup>	
CN legis levy (active made)	2.7 V ≤ V <sub>I</sub> ≤ 10 V		25°C			0.5	V	
EN logic low (active mode)	$ 2.7 \lor 5 \lor  5 10 \lor$		-40°C to 125°C			0.5	ı v	
EN hysteresis voltage			25°C		50		mV	
EN 's and some of	0 \ < \ \ < 10 \ \		25°C	-0.5	0.001	0.5		
EN input current	0 V ≤ V <sub>I</sub> ≤ 10 V		-40°C to 125°C	-0.5		0.5	μΑ	
Minimum V. for active page element			25°C		2.05	2.5	V	
Minimum V <sub>I</sub> for active pass element			-40°C to 125°C			2.5		
Minimum V. for volid DESET	10 (2000) - 200	A	25°C		1	1.5	V	
Minimum V <sub>I</sub> for valid RESET	IO(RESET) = -300	<i>μ</i> Α	-40°C to 125°C			1.9	l v	

<sup>‡</sup> CSR (compensation series resistance) refers to the total series resistance, including the equivalent series resistance (ESR) of the capacitor, any series resistance added externally, and PWB trace resistance to C<sub>0</sub>.

<sup>§</sup> Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.

SLVS124F – JUNE 1995 – REVISED JANUARY 1999

## TPS7301Q electrical characteristics at I<sub>O</sub> = 10 mA, V<sub>I</sub> = 3.5 V, $\overline{\text{EN}}$ = 0 V, C<sub>O</sub> = 4.7 $\mu\text{F}$ (CSR<sup>†</sup> = 1 $\Omega$ ), FB shorted to OUT at device leads (unless otherwise noted)

PARAMETER	TEST CO	ONDITIONS <sup>‡</sup>	TJ	MIN	TYP	MAX	UNIT	
			25°C		1.182		V	
Reference voltage (measured at FB)	$2.5 \text{ V} \leq \text{V}_{\text{I}} \leq 10 \text{ V},$ See Note 1	$5 \text{ mA} \le I_O \le 500 \text{ mA},$	-40°C to 125°C	1.147		1.217	V	
Reference voltage temperature coefficient			-40°C to 125°C		61	75	ppm/°C	
	V <sub>I</sub> = 2.4 V,	50 μA ≤ I <sub>O</sub> ≤ 150 mA	25°C		0.7	1		
	V  = 2.4 V,	30 μΑ 3 10 3 130 111Α	-40°C to 125°C			1		
	V <sub>I</sub> = 2.4 V,	150 mA ≤ I <sub>O</sub> ≤ 500 mA	25°C		0.83	1.3		
Pass-element series resistance	V  - 2.4 V,	130 IIIV 7 IO 7 200 IIIV	-40°C to 125°C			1.3	Ω	
(See Note 2)	V <sub>I</sub> = 2.9 V,	50 μA ≤ I <sub>O</sub> ≤ 500 mA	25°C		0.52	0.85	52	
	V  = 2.5 V,	30 μΑ 3 10 3 300 111Α	-40°C to 125°C			0.85		
	$V_{I} = 3.9 V,$	$50~\mu\text{A} \leq I_O \leq 500~\text{mA}$	25°C		0.32			
	$V_{ } = 5.9 V,$	$50~\mu\text{A} \leq I_O \leq 500~\text{mA}$	25°C		0.23			
Input regulation	$V_{ } = 2.5 \text{ V to } 10 \text{ V},$	$50~\mu\text{A} \leq \text{I}_{\mbox{O}} \leq 500~\text{mA},$	25°C		3	18	mV	
Imput regulation	See Note 1		-40°C to 125°C			25	111.0	
Output regulation	$2.5 \text{ V} \le \text{V}_{\text{I}} \le 10 \text{ V},$	$I_0 = 5 \text{ mA to } 500 \text{ mA},$	25°C		5	14	mV	
	See Note 1		-40°C to 125°C			25	111 V	
Output regulation			25°C		7	22	-l m\/ l	
	See Note 1		-40°C to 125°C			54		
		I <sub>O</sub> = 50 μA	25°C	48	59			
Ripple rejection	f = 120 Hz	ΙΟ = 30 μΑ	-40°C to 125°C	44		3 18 m' 25 5 14 m' 25 7 22 m' 54 59 de 54		
Kipple rejection	1 = 120112	I <sub>O</sub> = 500 mA,	25°C	45	54		l ub	
		See Note 1	-40°C to 125°C	44				
Output noise-spectral density	f = 120 Hz		25°C		2		μV/√ <del>Hz</del>	
		$C_0 = 4.7  \mu F$	25°C		95			
Output noise voltage	10 Hz ≤ f ≤ 100 kHz	C <sub>O</sub> = 10 μF	25°C		89		μVrms	
		C <sub>O</sub> = 100 μF	25°C		74			
RESET trip-threshold voltage§	V <sub>O(FB)</sub> decreasing		-40°C to 125°C	1.101		1.145	V	
RESET hysteresis voltage§	Measured at V <sub>O(FB)</sub>		25°C		12		mV	
DECET	Ì	1 100 :	25°C		0.1	0.4	.,	
RESET output low voltage§	V <sub>I</sub> = 2.13 V,	$IO(RESET) = 400 \mu A$	-40°C to 125°C		-	0.4	V	
ED insut comment			25°C	-10	0.1	10		
FB input current			-40°C to 125°C	-20		20	nA	

<sup>†</sup> CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to C<sub>0</sub>.

NOTES: 1. When V<sub>I</sub> < 2.9 V and I<sub>O</sub> > 150 mA simultaneously, pass element r<sub>DS(on)</sub> increases (see Figure 33) to a point where the resulting dropout voltage prevents the regulator from maintaining the specified tolerance range.

2. To calculate dropout voltage, use equation: VDO = IO · rDS(on) rDS(on) is a function of both output current and input voltage. This parametric table lists rDS(on) for VI = 2.4 V, 2.9 V, 3.9 V, and 5.9 V, which corresponds to dropout conditions for programmed output voltages of 2.5 V, 3 V, 4 V, and 6 V respectively. For other programmed values, refer to Figure 33.



<sup>‡</sup> Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.

<sup>§</sup> Output voltage programmed to 2.5 V with closed-loop configuration (see application information).

SLVS124F - JUNE 1995 - REVISED JANUARY 1999

# TPS7325Q electrical characteristics at I $_{O}$ = 10 mA, V $_{I}$ = 3.5 V, $\overline{EN}$ = 0 V, C $_{o}$ = 10 $\mu F$ (CSR $^{\dagger}$ = 1 $\Omega$ ), SENSE shorted to OUT (unless otherwise noted)

PARAMETER	TEST CO	NDITIONS <sup>‡</sup>	TJ	MIN	TYP	MAX	UNIT
Output voltage			25°C	2.45	2.5	2.55	V
Output voltage	$3.5 \text{ V} \le \text{V}_{\text{I}} \le 10 \text{ V},$	$5~\text{mA} \leq I_{\mbox{O}} \leq 500~\text{mA}$	-40°C to 125°C	2.425		2.575	V
	IO = 10 mA,	V <sub>I</sub> = 2.97 V	25°C		5		
	10 = 10 mA,	V  = 2.97 V	-40°C to 125°C			14	
Dropout voltage§	I <sub>O</sub> = 100 mA,	V <sub>I</sub> = 2.97 V	25°C		50	80	mV
Dropout voltages	10 = 100 mz,	V   - 2.57 V	-40°C to 125°C			150	) ''''
	IO = 500 mA,	V <sub>I</sub> = 2.97 V	25°C		270	400	
	10 = 300 IIIA,	V  = 2.97 V	-40°C to 125°C			600	
Pass-element series resistance§	(2.97 V – V <sub>O</sub> )/I <sub>O</sub> ,	V <sub>I</sub> = 2.97 V,	25°C		0.5	0.7	Ω
r ass-element series resistances	I <sub>O</sub> = 500 mA		-40°C to 125°C			1.4	22
nput regulation	V <sub>I</sub> = 3.5 V to 10 V,	50 μA ≤ I <sub>O</sub> ≤ 500 mA	25°C		6	6 20	mV
input regulation	V = 3.5 V to 10 V,	30 μΑ ≤ 10 ≤ 300 πΑ	-40°C to 125°C			25	IIIV
Output regulation	I <sub>O</sub> = 5 mA to 500 mA,	35 V < V, < 10 V	25°C		20	32	mV
	10 = 3 mA to 300 mA,	3.5 V \( \( \) \( \) \( \)	-40°C to 125°C			50	IIIV
Output regulation	$I_{O} = 50 \mu\text{A} \text{ to } 500 \text{mA},$	35 V < V1 < 10 V	25°C		28	60	mV
	10 = 30 μλ 10 300 πλ,	3.5 V 2 V 2 10 V	-40°C to 125°C		50 28 60 100	IIIV	
		ΙΟ = 50 μΑ	25°C	50	53		
Ripple rejection	f = 120 Hz	ΙΟ = 30 μΑ	-40°C to 125°C	49			dB
Kippie rejection	1 = 120112	10 - 500 mA	25°C	49	53		uБ
		I <sub>O</sub> = 500 mA	-40°C to 125°C	32			
Output noise-spectral density	f = 120 Hz		25°C		2		μV/√ <del>Hz</del>
		$C_0 = 4.7  \mu F$	25°C		274		
Output noise voltage	10 Hz ≤ f ≤ 100 kHz	C <sub>O</sub> = 10 μF	25°C		228		μVrms
		C <sub>O</sub> = 100 μF	25°C		159		
RESET trip-threshold voltage	V <sub>O</sub> decreasing	•	-40°C to 125°C	2.23	2.32	2.39	V
	V 04V		25°C		0.14	0.4	
RESET output low voltage	V <sub>I</sub> = 2.1 V,	IO(RESET) = -0.8  mA	-40°C to 125°C			0.4	V

TCSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to Co.

<sup>‡</sup> Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.

<sup>§</sup> Dropout test and pass-element series resistance test are not production tested. Test method requires SENSE terminal to be disconnected from output voltage.

SLVS124F – JUNE 1995 – REVISED JANUARY 1999

# TPS7330Q electrical characteristics at I $_{O}$ = 10 mA, V $_{I}$ = 4 V, $\overline{EN}$ = 0 V, C $_{O}$ = 4.7 $\mu\text{F}$ (CSR $^{\dagger}$ = 1 $\Omega$ ), SENSE shorted to OUT (unless otherwise noted)

PARAMETER	TEST CO	NDITIONS <sup>‡</sup>	TJ	MIN	TYP	MAX	UNIT
Custout valtage			25°C		3		V
Output voltage	$4 \text{ V} \le V_{I} \le 10 \text{ V},$	$5 \text{ mA} \le I_O \le 500 \text{ mA}$	-40°C to 125°C	2.94		3.06	V
	IO = 10 mA,	V <sub>I</sub> = 2.94 V	25°C		5.2	7	
	IO = 10 IIIA,	V  = 2.94 V	-40°C to 125°C			10	
Dropout voltage	I <sub>O</sub> = 100 mA,	V <sub>I</sub> = 2.94 V	25°C		52	75	mV
Dropout voltage	10 = 100 mz,	V   = 2.94 V	-40°C to 125°C	100	111 V		
	IO = 500 mA,	V <sub>I</sub> = 2.94 V	25°C		267	450	]
	10 = 300 mA,	V  = 2.94 V	-40°C to 125°C			500	
Pass-element series resistance	(2.94 V – V <sub>O</sub> )/I <sub>O</sub> ,	V <sub>I</sub> = 2.94 V,	25°C		0.5	0.7	Ω
Fass-element series resistance	$I_{O} = 500 \text{ mA}$		-40°C to 125°C			1	22
Input regulation	V <sub>I</sub> = 4 V to 10 V,	50 μA ≤ I <sub>O</sub> ≤ 500 mA	25°C		6	3 23	mV
Input regulation	V = 4 V 10 10 V,	20 ha ≥ 10 ≥ 200 ma	-40°C to 125°C			29	IIIV
Output regulation	I <sub>O</sub> = 5 mA to 500 mA,	4 V < V < 10 V	25°C		20	32	mV
	10 = 3 mA to 300 mA,	4 V \( \times V \( \) \( \) \( \)	-40°C to 125°C			60	IIIV
Output regulation	$I_{O} = 50 \mu\text{A} \text{ to } 500 \text{mA},$	4 V < V1 < 10 V	25°C		28	60	mV
	$10 = 30 \mu\text{A to } 300 \text{mA},$	4 0 2 0 2 10 0	-40°C to 125°C			120	IIIV
		ΙΟ = 50 μΑ	25°C	43	53		
Ripple rejection	f = 120 Hz	ΙΟ = 50 μΑ	-40°C to 125°C	40			dB
Ripple rejection	T = 120 HZ	10 - 500 mA	25°C	39	53		uБ
		I <sub>O</sub> = 500 mA	-40°C to 125°C	36			
Output noise-spectral density	f = 120 Hz		25°C		2		μV/√ <del>Hz</del>
		$C_0 = 4.7  \mu F$	25°C		274		
Output noise voltage	10 Hz ≤ f ≤ 100 kHz	C <sub>O</sub> = 10 μF	25°C		228		μVrms
		C <sub>O</sub> = 100 μF	25°C		159		
RESET trip-threshold voltage	V <sub>O</sub> decreasing		-40°C to 125°C	2.58	2.64	2.7	V
DEOET autout laws alto as	V 00V		25°C		0.14	0.4	.,
RESET output low voltage	V <sub>I</sub> = 2.6 V,	IO(RESET) = -0.8  mA	-40°C to 125°C		•	0.4	V

<sup>†</sup> CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to Co.

<sup>‡</sup> Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.

SLVS124F - JUNE 1995 - REVISED JANUARY 1999

# TPS7333Q electrical characteristics at I $_{O}$ = 10 mA, V $_{I}$ = 4.3 V, $\overline{EN}$ = 0 V, C $_{O}$ = 4.7 $\mu\text{F}$ (CSR $^{\dagger}$ = 1 $\Omega$ ), SENSE shorted to OUT (unless otherwise noted)

PARAMETER	TEST CO	NDITIONS <sup>‡</sup>	TJ	MIN	TYP	MAX	UNIT
Outrot walta na			25°C		3.3		V
Output voltage	$4.3 \text{ V} \le \text{V}_{\text{I}} \le 10 \text{ V},$	$5 \text{ mA} \le I_{O} \le 500 \text{ mA}$	-40°C to 125°C	3.23		3.37	V
	10 mA	V. 2.22.V	25°C		4.5	7	
	$I_O = 10 \text{ mA},$	V <sub>I</sub> = 3.23 V	-40°C to 125°C			8	
Description	In - 100 mA	V. – 2 22 V	25°C		44	60	mV
Dropout voltage	$I_O = 100 \text{ mA},$	V <sub>I</sub> = 3.23 V	-40°C to 125°C			80	IIIV
	IO = 500 mA,	V <sub>1</sub> = 2.22 V	25°C		235	300	
	IQ = 500 IIIA,	$V_{\parallel} = 3.23 \text{ V}$ $-40^{\circ}\text{C to}$				400	
Pass-element series resistance	(3.23 V – V <sub>O</sub> )/I <sub>O</sub> ,	V <sub>I</sub> = 3.23 V,	25°C		0.44	0.6	Ω
Pass-element series resistance	I <sub>O</sub> = 500 mA	•	-40°C to 125°C			0.8	22
lanut va quilatian	V: 42 V to 40 V	50 A < la < 500 mA	25°C		6 23	mV	
Input regulation	$V_{I} = 4.3 \text{ V to } 10 \text{ V},$	$50 \mu\text{A} \le I_{O} \le 500 \text{mA}$	-40°C to 125°C			29	IIIV
	I <sub>O</sub> = 5 mA to 500 mA	421/21/2101/	25°C		21	38	mV
Output regulation	10 = 2 HIM 10 200 HIM	$  4.3 \text{ V} \leq \text{V}   \leq 10 \text{ V}$	-40°C to 125°C			75	IIIV
	Io - 50 uA to 500 m/	$I_O = 50 \mu A$ to 500 mA, 4.3 V $\leq V_I \leq 10 \text{ V}$			31	60	mV
	10 = 50 μΑ (0 500 ΠΙΑ	$A, 4.5 \lor 5 \lor   5 \lor 0 \lor$	-40°C to 125°C			120	IIIV
		ΙΟ = 50 μΑ	25°C	43	51		
Ripple rejection	f = 120 Hz	ΙΟ = 50 μΑ	-40°C to 125°C	40			dB
Rippie rejection	T = 120 HZ	IO = 500 mA	25°C	39	49		uБ
		IQ = 500 IIIA	-40°C to 125°C	36			
Output noise-spectral density	f = 120 Hz		25°C		2		μV/√Hz
		$C_0 = 4.7  \mu F$	25°C		274		
Output noise voltage	10 Hz ≤ f ≤ 100 kHz	C <sub>O</sub> = 10 μF	25°C		228		μVrms
		C <sub>O</sub> = 100 μF	25°C		159		
RESET trip-threshold voltage	V <sub>O</sub> decreasing		-40°C to 125°C	2.868			V
RESET hysteresis voltage			25°C		18		mV
RESET output low voltage	V. 20V	1	25°C		0.17	0.4	
KESET output low voitage	V <sub>I</sub> = 2.8 V,	IO(RESET) = -1 mA	-40°C to 125°C			0.4	V

<sup>†</sup>CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to C<sub>O</sub>.

<sup>&</sup>lt;sup>‡</sup> Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.

SLVS124F – JUNE 1995 – REVISED JANUARY 1999

# TPS7348Q electrical characteristics at I<sub>O</sub> = 10 mA, V<sub>I</sub> = 5.85 V, $\overline{\text{EN}}$ = 0 V, C<sub>O</sub> = 4.7 $\mu\text{F}$ (CSR<sup>†</sup> = 1 $\Omega$ ), SENSE shorted to OUT (unless otherwise noted)

PARAMETER	TEST CO	NDITIONS <sup>‡</sup>	TJ	MIN	TYP	MAX	UNIT
Output voltogo			25°C		4.85		V
Output voltage	$5.85 \text{ V} \le \text{V}_{I} \le 10 \text{ V},$	$5 \text{ mA} \le I_O \le 500 \text{ mA}$	-40°C to 125°C	4.75		4.95	V
	I <sub>O</sub> = 10 mA,	V <sub>I</sub> = 4.75 V	25°C		2.9	6	
	IO = 10 IIIA,	V  = 4.75 V	-40°C to 125°C			8	
Dropout voltogo	I <sub>O</sub> = 100 mA,	V <sub>I</sub> = 4.75 V	25°C		28	37	mV
Dropout voltage	10 = 100 IIIA,	V   = 4.75 V	-40°C to 125°C			54	IIIV
	I <sub>O</sub> = 500 mA,	V <sub>I</sub> = 4.75 V	25°C		150	180	
	10 = 300 IIIA,	V   = 4.75 V	-40°C to 125°C			250	
Pass-element series resistance	(4.75 V – V <sub>O</sub> )/I <sub>O</sub> ,	V <sub>I</sub> = 4.75 V,	25°C		0.28	0.37	Ω
r ass-element series resistance	I <sub>O</sub> = 500 mA		-40°C to 125°C			0.52	52
Input regulation	\/\. = 5.95.\/ to 10.\/	50 μA ≤ I <sub>O</sub> ≤ 500 mA	25°C		9	35	mV
Imput regulation	V  = 5.65 V to 10 V,	50 μA ≤ 1O ≤ 500 IIIA	-40°C to 125°C			37	IIIV
	I <sub>O</sub> = 5 mA to 500 mA	5 95 V < V, < 10 V	25°C		28	42	mV
Output regulation	10 = 3 IIIA to 300 IIIA	, 5.85 V \(\sigma\) \(\sigma\)	-40°C to 125°C			80	IIIV
	$I_O = 50 \mu A \text{ to } 500 \text{ mA}, 5.85 \text{ V} \le V_I \le 10 \text{ V}$		25°C		42	65	mV
	10 = 30 μΑ 10 300 1114	x, 5.85 v ≤ v  ≤ 10 v	-40°C to 125°C			130	mv
		ΙΟ = 50 μΑ	25°C	42	53		
Pipple rejection	f = 120 Hz	10 = 30 μΑ	-40°C to 125°C	39			dB
Rippie rejection	1 = 120112	I <sub>O</sub> = 500 mA	25°C	39	50		uБ
Ripple rejection		IQ = 300 IIIA	-40°C to 125°C	35			
Output noise-spectral density	f = 120 Hz		25°C		2		μV/√Hz
		$C_0 = 4.7  \mu F$	25°C		410		
Output noise voltage	10 Hz ≤ f ≤ 100 kHz	C <sub>O</sub> = 10 μF	25°C		328		μVrms
		C <sub>O</sub> = 100 μF	25°C		212		
RESET trip-threshold voltage	V <sub>O</sub> decreasing		-40°C to 125°C	4.5		4.7	V
RESET hysteresis voltage			25°C		26		mV
DECET cutput lovu valtage	1	A V. 440 V	25°C		0.2	0.4	V
RESET output low voltage	IO(RESET) = -1.2  m	A, V   = 4.12 V	-40°C to 125°C			0.4	

<sup>†</sup> CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to C<sub>0</sub>.

<sup>‡</sup> Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.

SLVS124F - JUNE 1995 - REVISED JANUARY 1999

# TPS7350Q electrical characteristics at I $_{O}$ = 10 mA, V $_{I}$ = 6 V, $\overline{EN}$ = 0 V, C $_{O}$ = 4.7 $\mu\text{F}$ (CSR $^{\dagger}$ = 1 $\Omega$ ), SENSE shorted to OUT (unless otherwise noted)

PARAMETER	TEST CON	IDITIONS <sup>‡</sup>	TJ	MIN	TYP	MAX	UNIT	
Output with in			25°C		5		V	
Output voltage	6 V ≤ V <sub>I</sub> ≤ 10 V,	$5 \text{ mA} \le I_O \le 500 \text{ mA}$	-40°C to 125°C	4.9		5.1	l <sup>v</sup>	
	1- 10 mA	\/ <sub>2</sub>	25°C		2.9	6		
	I <sub>O</sub> = 10 mA,	V <sub>I</sub> = 4.88 V	-40°C to 125°C			8		
Dropout voltogo	IO = 100 mA,	V <sub>I</sub> = 4.88 V	25°C		27	35	mV	
Dropout voltage	10 = 100 mA,	V = 4.00 V	-40°C to 125°C			50	1111	
	I <sub>O</sub> = 500 mA,	V <sub>I</sub> = 4.88 V	25°C		146	170		
	10 = 300 mA,	V  = 4.00 V	-40°C to 125°C			230		
Pass-element series resistance	(4.88 V – V <sub>O</sub> )/I <sub>O</sub> ,	V <sub>I</sub> = 4.88 V,	25°C		0.27	0.35	Ω	
Pass-element series resistance	$I_{O} = 500 \text{ mA}$		-40°C to 125°C			0.5	52	
Input regulation	V <sub>I</sub> = 6 V to 10 V,	$50  \mu A \le I_O \le 500  mA$	25°C		4	25	mV	
Input regulation	V = 6 V 10 10 V,	20 μA ≥ 1Ω ≥ 200 IIIA	-40°C to 125°C			45	IIIV	
	$I_{O} = 5 \text{ mA to } 500 \text{ mA},$	6 \/ < \/ < 10 \/	25°C		30	45	mV	
Output regulation	10 = 3 mA to 300 mA,	0 v ≥ v  ≥ 10 v	-40°C to 125°C			86	IIIV	
	lo - 50 u \ to 500 m \	nA, 6 V ≤ V <sub>I</sub> ≤ 10 V	25°C		45	65	mV	
	ΙΟ = 50 μΑ το 500 ΠΑ,	0 v ≥ v  ≥ 10 v	-40°C to 125°C			140	IIIV	
		ΙΟ = 50 μΑ	25°C	43	53			
Ripple rejection	f = 120 Hz	ΙΟ = 30 μΑ	-40°C to 125°C	38			dB	
Ripple rejection	1 = 120 HZ	I <sub>O</sub> = 500 mA	25°C	41	51		uБ	
		IQ = 200 IIIA	-40°C to 125°C	36				
Output noise-spectral density	f = 120 Hz		25°C		2		μV/√ <del>Hz</del>	
		$C_0 = 4.7  \mu F$	25°C		430			
Output noise voltage	10 Hz ≤ f ≤ 100 kHz	C <sub>O</sub> = 10 μF	25°C		345		μVrms	
		C <sub>O</sub> = 100 μF	25°C		220			
RESET trip-threshold voltage	V <sub>O</sub> decreasing	•	-40°C to 125°C	4.55		4.75	V	
RESET hysteresis voltage			25°C		28		mV	
DECET autaut lauvualtaaa		V 405 V	25°C		0.15	0.4	.,	
RESET output low voltage	IO(RESET) = -1.2  mA	, VI = 4.25 V	-40°C to 125°C			0.4	V	

<sup>†</sup> CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to C<sub>0</sub>.

<sup>&</sup>lt;sup>‡</sup> Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.

SLVS124F – JUNE 1995 – REVISED JANUARY 1999

### switching characteristics

PARAMETER	TEST CONDITIONS		TPS7301 TPS7348	,		UNIT
			MIN	TYP	MAX	
DESET time out delay	Coo Figure F	25°C	140	200	260	
RESET time-out delay	See Figure 5	-40°C to 125°C	100		300	ms

# electrical characteristics at I $_{O}$ = 10 mA, $\overline{EN}$ = 0 V, C $_{O}$ = 4.7 $\mu F$ (CSR $^{\dagger}$ = 1 $\Omega$ ), T $_{J}$ = 25°C, SENSE/FB shorted to OUT (unless otherwise noted)

PARAMETER	TEST CON	TEST CONDITIONS‡			TPS7301Y, TPS7333Y TPS7348Y, TPS7350Y		
			MIN	TYP	MAX		
Ground current (active mode)	$\overline{\text{EN}} \le 0.5 \text{ V},$ 0 mA $\le \text{I}_{\text{O}} \le 500 \text{ mA}$	$V_I = V_O + 1 V$ ,		340		μА	
Input current (standby mode)	$\overline{EN} = V_{I},$	$2.7~\text{V} \leq \text{V}_{\text{I}} \leq 10~\text{V}$		0.01		μΑ	
Output current limit	$V_0 = 0 V$ ,	V <sub>I</sub> = 10 V		1.2		Α	
Pass-element leakage current in standby mode	$\overline{EN} = V_{I},$	$2.7~V \leq V_{\mbox{\scriptsize I}} \leq 10~V$		0.01		μΑ	
RESET leakage current	Normal operation,	V at RESET = 10 V		0.02		μΑ	
Thermal shutdown junction temperature				165		°C	
EN logic low (active mode)	$2.7 \text{ V} \le \text{V}_{\text{I}} \le 10 \text{ V}$				0.5	V	
EN hysteresis voltage				50		mV	
EN input current	0 V ≤ V <sub>I</sub> ≤ 10 V			0.001		μΑ	
Minimum V <sub>I</sub> for active pass element				2.05		V	
Minimum V <sub>I</sub> for valid RESET	$IO(RESET) = -300 \mu$	A		1	·	V	

<sup>†</sup> CSR (compensation series resistance) refers to the total series resistance, including the equivalent series resistance (ESR) of the capacitor, any series resistance added externally, and PWB trace resistance to C<sub>0</sub>.



<sup>‡</sup> Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.

SLVS124F - JUNE 1995 - REVISED JANUARY 1999

## TPS7301Y electrical characteristics at I<sub>O</sub> = 10 mA, V<sub>I</sub> = 3.5 V, $\overline{\text{EN}}$ = 0 V, C<sub>O</sub> = 4.7 $\mu\text{F}$ (CSR<sup>†</sup> = 1 $\Omega$ ), T<sub>J</sub> = 25°C, FB shorted to OUT at device leads (unless otherwise noted)

PARAMETER	TEST CO	ONDITIONS <sup>‡</sup>	MIN TYP	MAX	UNIT
Reference voltage (measured at FB)			1.182		V
	V <sub>I</sub> = 2.4 V,	$50 \ \mu\text{A} \le I_{\mbox{O}} \le 150 \ \mbox{mA}$	0.7		
	V <sub>I</sub> = 2.4 V,	$150~\text{mA} \leq I_{\mbox{O}} \leq 500~\text{mA}$	0.83		
Pass-element series resistance (See Note 2)	V <sub>I</sub> = 2.9 V,	$50 \ \mu A \le I_O \le 500 \ mA$	0.52		Ω
	V <sub>I</sub> = 3.9 V,	$50 \ \mu\text{A} \le I_{\mbox{O}} \le 500 \ \mbox{mA}$	0.32		
	V <sub>I</sub> = 5.9 V,	$50 \ \mu\text{A} \le I_{\mbox{O}} \le 500 \ \mbox{mA}$	0.23		
Input regulation	V <sub>I</sub> = 2.5 V to 10 V, See Note 1	$50 \mu A \le I_O \le 500 \text{ mA},$	3		mV
Outros de mandation	$2.5 \text{ V} \le \text{V}_{\text{I}} \le 10 \text{ V},$ See Note 1	$I_O = 5$ mA to 500 mA,	5		mV
Output regulation	$2.5 \text{ V} \le \text{V}_{\text{I}} \le 10 \text{ V},$ See Note 1	$I_O = 50 \mu\text{A} \text{ to } 500 \text{mA},$	7		mV
		ΙΟ = 50 μΑ	59		
Ripple rejection	f = 120 Hz	I <sub>O</sub> = 500 mA, See Note 1	54		dB
Output noise-spectral density	f = 120 Hz		2		μV/√Hz
		$C_0 = 4.7 \mu\text{F}$	95		
Output noise voltage	10 Hz ≤ f ≤ 100 kHz	C <sub>O</sub> = 10 μF	89		μVrms
		C <sub>O</sub> = 100 μF	74		
RESET hysteresis voltage§	Measured at VO(FB)		12		mV
RESET output low voltage§	V <sub>I</sub> = 2.13 V,	I <sub>O</sub> (RESET) = 400 μA	0.1		V
FB input current			0.1		nA

<sup>†</sup> CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to Co.

To calculate dropout voltage, use equation: V<sub>DO</sub> = I<sub>O</sub> · r<sub>DS</sub>(on) r<sub>DS</sub>(on) is a function of both output current and input voltage. The parametric table lists r<sub>DS</sub>(on) for V<sub>I</sub> = 2.4 V, 2.9 V, 3.9 V, and 5.9 V, which corresponds to dropout conditions for programmed output voltages of 2.5 V, 3 V, 4 V, and 6 V respectively. For other programmed values, refer to Figure 33.

<sup>‡</sup> Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.

<sup>§</sup> Output voltage programmed to 2.5 V with closed-loop configuration (see application information).

NOTES: 1. When V<sub>I</sub> < 2.9 V and I<sub>O</sub> > 150 mA simultaneously, pass element r<sub>DS(On)</sub> increases (see Figure 33) to a point where the resulting dropout voltage prevents the regulator from maintaining the specified tolerance range.

SLVS124F – JUNE 1995 – REVISED JANUARY 1999

## TPS7325Y electrical characteristics at I<sub>O</sub> = 10 mA, V<sub>I</sub> = 3.5 V, $\overline{\text{EN}}$ = 0 V, C<sub>O</sub> = 10 $\mu\text{F}$ (CSR<sup>†</sup> = 1 $\Omega$ ), T<sub>J</sub> = 25°C, SENSE shorted to OUT (unless otherwise noted)

PARAMETER	TEST COI	TEST CONDITIONS <sup>‡</sup>		MAX	UNIT
Output voltage			2.5		V
	I <sub>O</sub> = 10 mA,	V <sub>I</sub> = 2.97 V	5		
Dropout voltage§	$I_0 = 100 \text{ mA},$	V <sub>I</sub> = 2.97 V	50		mV
	$I_0 = 500 \text{ mA},$	V <sub>I</sub> = 2.97 V	270		
Pass-element series resistance§	(2.97 V – V <sub>O</sub> )/I <sub>O</sub> , I <sub>O</sub> = 500 mA	V <sub>I</sub> = 2.97 V,	0.5		Ω
Input regulation	$V_{I} = 3.5 \text{ V to } 10 \text{ V},$	$50 \mu\text{A} \le I_{\mbox{O}} \le 500  \mbox{mA}$	6		mV
	$I_O = 5 \text{ mA to } 500 \text{ mA},$	3.5 V ≤ V <sub>I</sub> ≤ 10 V	20		mV
Output regulation	$I_{O} = 50 \mu A \text{ to } 500 \text{ mA},$	3.5 V ≤ V <sub>I</sub> ≤ 10 V	28		mV
Dinnle rejection	f = 120 Hz	ΙΟ = 50 μΑ	53		dB
Ripple rejection	1 = 120 Hz	I <sub>O</sub> = 500 mA	53		uБ
Output noise-spectral density	f = 120 Hz		2		μV/√ <del>Hz</del>
		$C_0 = 4.7  \mu F$	274		
Output noise voltage	10 Hz ≤ f ≤ 100 kHz	C <sub>O</sub> = 10 μF	228		μVrms
		C <sub>O</sub> = 100 μF	159		
RESET output low voltage	V <sub>I</sub> = 2.1 V,	$I_{O(RESET)} = -0.8 \text{ mA}$	0.14		V

<sup>†</sup> CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to Co.



<sup>‡</sup> Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.

<sup>§</sup> Dropout test and pass-element series resistance test are not production tested. Test method requires SENSE terminal to be disconnected from output voltage.

SLVS124F - JUNE 1995 - REVISED JANUARY 1999

## TPS7330Y electrical characteristics at I<sub>O</sub> = 10 mA, V<sub>I</sub> = 4 V, $\overline{\text{EN}}$ = 0 V, C<sub>O</sub> = 4.7 $\mu\text{F}$ (CSR<sup>†</sup> = 1 $\Omega$ ), T<sub>J</sub> = 25°C, SENSE shorted to OUT (unless otherwise noted)

PARAMETER	TEST COI	TEST CONDITIONS <sup>‡</sup>		MAX	UNIT
Output voltage			3		V
	I <sub>O</sub> = 10 mA,	V <sub>I</sub> = 2.94 V	5.2		
Dropout voltage	I <sub>O</sub> = 100 mA,	V <sub>I</sub> = 2.94 V	52		mV
	$I_0 = 500 \text{ mA},$	V <sub>I</sub> = 2.94 V	267		
Pass-element series resistance	$(2.94 \text{ V} - \text{V}_{\text{O}})/\text{I}_{\text{O}},$ $\text{I}_{\text{O}} = 500 \text{ mA}$	V <sub>I</sub> = 2.94 V,	0.5		Ω
Input regulation	$V_{I} = 4 \text{ V to } 10 \text{ V},$	$50 \mu\text{A} \le I_{O} \le 500 \text{mA}$	6		mV
Output regulation	$I_O = 5 \text{ mA to } 500 \text{ mA},$	4 V ≤ V <sub>I</sub> ≤ 10 V	20		mV
Output regulation	$I_0 = 50 \mu A \text{ to } 500 \text{ mA},$	4 V ≤ V <sub>I</sub> ≤ 10 V	28		mV
Dinale rejection	f = 120 Hz	ΙΟ = 50 μΑ	53		dB
Ripple rejection	I = 120 HZ	I <sub>O</sub> = 500 mA	53		uв
Output noise-spectral density	f = 120 Hz		2		μV/√ <del>Hz</del>
		$C_0 = 4.7 \mu\text{F}$	274		
Output noise voltage	10 Hz ≤ f ≤ 100 kHz	C <sub>O</sub> = 10 μF	228		μVrms
		C <sub>O</sub> = 100 μF	159		
RESET output low voltage	V <sub>I</sub> = 2.6 V,	$I_{O(RESET)} = -0.8 \text{ mA}$	0.14		V

<sup>†</sup> CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to Co.

## TPS7333Y electrical characteristics at I<sub>O</sub> = 10 mA, V<sub>I</sub> = 4.3 V, $\overline{\text{EN}}$ = 0 V, C<sub>O</sub> = 4.7 $\mu\text{F}$ (CSR<sup>†</sup> = 1 $\Omega$ ), T<sub>J</sub> = 25°C, SENSE shorted to OUT (unless otherwise noted)

PARAMETER	TEST CON	TEST CONDITIONS <sup>‡</sup>		MAX	UNIT
Output voltage			3.3		V
	I <sub>O</sub> = 10 mA,	V <sub>I</sub> = 3.23 V	4.5		
Dropout voltage	$I_O = 100 \text{ mA},$	V <sub>I</sub> = 3.23 V	44		mV
	$I_{O} = 500 \text{ mA},$	V <sub>I</sub> = 3.23 V	235		
Pass-element series resistance	$(3.23 \text{ V} - \text{V}_{\text{O}})/\text{I}_{\text{O}},$ $\text{I}_{\text{O}} = 500 \text{ mA}$	V <sub>I</sub> = 3.23 V,	0.44		Ω
Input regulation	$V_I = 4.3 \text{ V to } 10 \text{ V},$	$50  \mu A \le I_O \le 500  mA$	6		mV
Output regulation	$I_{O} = 5 \text{ mA to } 500 \text{ mA},$	4.3 V ≤ V <sub>I</sub> ≤ 10 V	21		mV
	$I_O = 50 \mu\text{A} \text{ to } 500 \text{mA},$	4.3 V ≤ V <sub>I</sub> ≤ 10 V	31		mV
Pinnla raination	f = 120 Hz	I <sub>O</sub> = 50 μA	51		dB
Ripple rejection	T = 120 HZ	I <sub>O</sub> = 500 mA	49		uБ
Output noise-spectral density	f = 120 Hz		2		μV/√ <del>Hz</del>
		$C_0 = 4.7 \mu\text{F}$	274		
Output noise voltage	10 Hz ≤ f ≤ 100 kHz	C <sub>O</sub> = 10 μF	228		μVrms
		C <sub>O</sub> = 100 μF	159		
RESET hysteresis voltage			18		mV
RESET output low voltage	V <sub>I</sub> = 2.8 V,	IO(RESET) = -1 mA	0.17		V

<sup>†</sup> CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to Co.

<sup>‡</sup> Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.



<sup>‡</sup> Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.

SLVS124F – JUNE 1995 – REVISED JANUARY 1999

# TPS7348Y electrical characteristics at I $_{O}$ = 10 mA, V $_{I}$ = 5.85 V, $\overline{EN}$ = 0 V, C $_{O}$ = 4.7 $\mu$ F (CSR $^{\dagger}$ = 1 $\Omega$ ), T $_{J}$ = 25°C, SENSE shorted to OUT (unless otherwise noted)

PARAMETER	TEST CON	iditions‡	MIN TYP	MAX	UNIT
Output voltage			4.85		V
	I <sub>O</sub> = 10 mA,	V <sub>I</sub> = 4.75 V	2.9		
Dropout voltage	$I_{O} = 100 \text{ mA},$	V <sub>I</sub> = 4.75 V	28		mV
	$I_{O} = 500 \text{ mA},$	V <sub>I</sub> = 4.75 V	150		
Pass-element series resistance	$(4.75 \text{ V} - \text{V}_{\text{O}})/I_{\text{O}},$ $I_{\text{O}} = 500 \text{ mA}$	V <sub>I</sub> = 4.75 V,	0.28		Ω
Input regulation	V <sub>I</sub> = 5.85 V to 10 V,	$50  \mu A \le I_O \le 500  mA$	9		mV
Outrot as relation	$I_O = 5 \text{ mA to } 500 \text{ mA},$	5.85 V ≤ V <sub>I</sub> ≤ 10 V	28		mV
Output regulation	$I_{O} = 50 \mu\text{A} \text{ to } 500 \text{mA},$	$5.85 \text{ V} \le \text{V}_{1} \le 10 \text{ V}$	42		mV
Ripple rejection	f = 120 Hz	ΙΟ = 50 μΑ	53		dB
Ripple rejection	I = 120 Hz	I <sub>O</sub> = 500 mA	50		uБ
Output noise-spectral density	f = 120 Hz		2		μV/√ <del>Hz</del>
		$C_0 = 4.7  \mu F$	410		
Output noise voltage	10 Hz ≤ f ≤ 100 kHz	C <sub>O</sub> = 10 μF	328		μVrms
		C <sub>O</sub> = 100 μF	212		
RESET hysteresis voltage		-	26		mV
RESET output low voltage	IO(RESET) = -1.2  mA,	V <sub>I</sub> = 4.12 V	0.2		V

<sup>†</sup> CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to C<sub>0</sub>.



<sup>‡</sup> Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.

SLVS124F - JUNE 1995 - REVISED JANUARY 1999

# TPS7350Y electrical characteristics at I $_{O}$ = 10 mA, V $_{I}$ = 6 V, $\overline{EN}$ = 0 V, C $_{O}$ = 4.7 $\mu F$ (CSR $^{\dagger}$ = 1 $\Omega$ ), T $_{J}$ = 25°C, SENSE shorted to OUT (unless otherwise noted)

PARAMETER	TEST CON	IDITIONS <sup>‡</sup>	MIN T	ΥP	MAX	UNIT
Output voltage				5		V
	I <sub>O</sub> = 10 mA,	V <sub>I</sub> = 4.88 V		2.9	6	
Dropout voltage	$I_{O} = 100 \text{ mA},$	V <sub>I</sub> = 4.88 V		27	35	mV
	$I_{O} = 500 \text{ mA},$	V <sub>I</sub> = 4.88 V		146	170	
Pass-element series resistance	$(4.88 \text{ V} - \text{V}_{\text{O}})/\text{I}_{\text{O}},$ $\text{I}_{\text{O}} = 500 \text{ mA}$	V <sub>I</sub> = 4.88 V,	0	.27	0.35	Ω
Input regulation	$V_{I} = 6 \text{ V to } 10 \text{ V},$	$50  \mu A \le I_O \le 500  mA$		4	25	mV
Outroit as mulation	$I_0 = 5 \text{ mA to } 500 \text{ mA},$	6 V ≤ V <sub>I</sub> ≤ 10 V		28	75	mV
Output regulation	$I_{O} = 50 \mu\text{A} \text{ to } 500 \text{mA},$	6 V ≤ V <sub>I</sub> ≤ 10 V		41		mV
Pinnle rejection	f = 120 Hz	ΙΟ = 50 μΑ		53		dB
Ripple rejection	1 = 120 Hz	I <sub>O</sub> = 500 mA		51		иь
Output noise-spectral density	f = 120 Hz			2		μV/√ <del>Hz</del>
		$C_0 = 4.7  \mu F$	4	430		
Output noise voltage	10 Hz ≤ f ≤ 100 kHz	C <sub>O</sub> = 10 μF	;	345		μVrms
		C <sub>O</sub> = 100 μF	2	220		
RESET hysteresis voltage				28		mV
RESET output low voltage	IO(RESET) = -1.2  mA,	V <sub>I</sub> = 4.25 V	0	.15	0.4	V

<sup>†</sup> CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to C<sub>0</sub>.



<sup>‡</sup> Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.

### PARAMETER MEASUREMENT INFORMATION

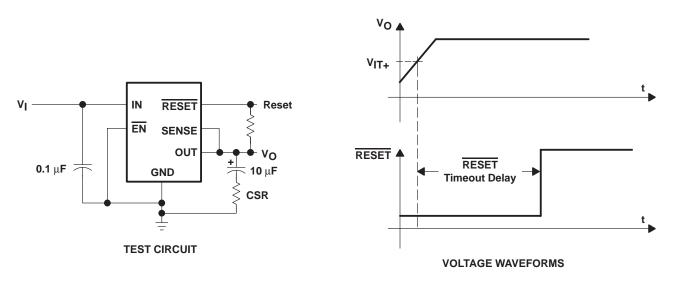


Figure 5. Test Circuit and Voltage Waveforms

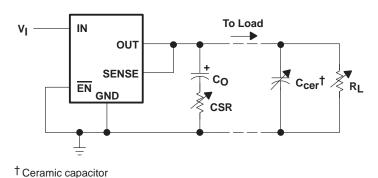


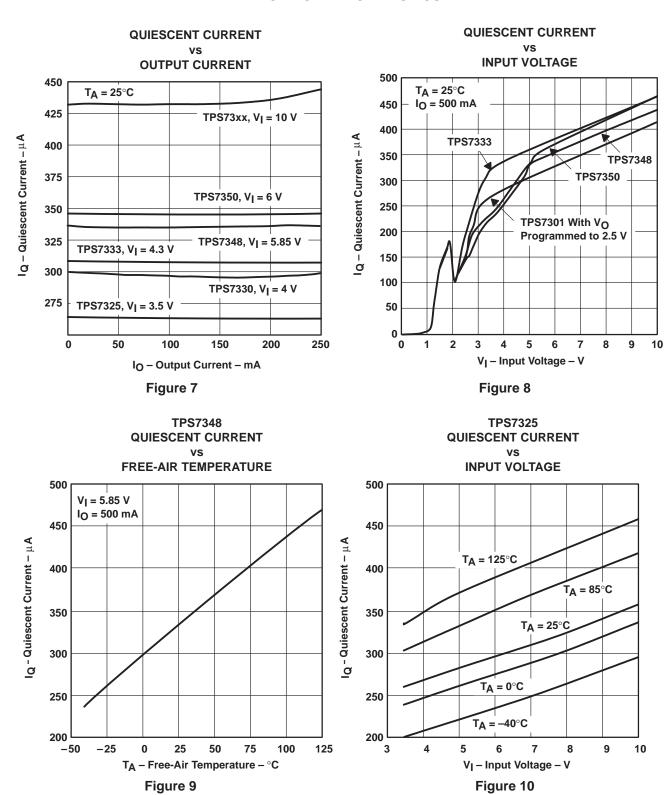
Figure 6. Test Circuit for Typical Regions of Stability (Refer to Figures 29 through 32)

SLVS124F - JUNE 1995 - REVISED JANUARY 1999

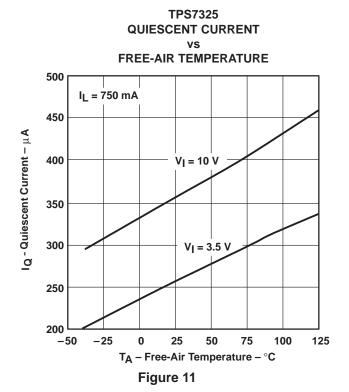
### **TYPICAL CHARACTERISTICS**

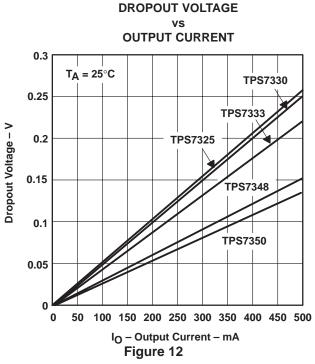
### **Table of Graphs**

			vs Output current	7
Q	Quiescent current		vs Input voltage	8
Q	Quiescent current	TPS7348	vs Free-air temperature	9
			vs Input voltage	10
Q	Quiescent current	TPS7325	vs Free-air temperature	11
√DO	Dropout voltage	•	vs Output current	12
7/DO	Change in dropout voltage		vs Free-air temperature	13
√DO	Dropout voltage	TPS7301	vs Output current	14
7^O	Change in output voltage	•	vs Free-air temperature	15
/o	Output voltage		vs Input voltage	16
/ <sub>O</sub>	Output voltage	TPS7325	vs Input voltage	17
	Line regulation	•		18
		TPS7301	vs Output current	19
	Output voltage	TPS7325	vs Output current	20
		TPS7330	vs Output current	21
Vo		TPS7333	vs Output current	22
		TPS7348	vs Output current	23
		TPS7350	vs Output current	24
	Output voltage response from enable	e (EN)		25
		TPS7301 or TPS7333		26
		TPS7325		27
	Load transient response	TPS7348 or TPS7350		28
	Load transient response	TPS7301		29
		TPS7333		30
		TPS7348 or TPS7350		31
	Ripple rejection		vs Frequency	32
	Output spectral noise density		vs Frequency	33
		C <sub>O</sub> = 4.7 μF	vs Output current	34
	Compensation series resistance	$C_0 = 4.7  \mu F$	vs Added ceramic capacitance	35
	(CSR)	C - 10 uE	vs Output current	36
		C <sub>O</sub> = 10 μF	vs Added ceramic capacitance	37
DS(on)	Pass-element resistance		vs Input voltage	38
V <sub>I</sub>	Minimum input voltage for valid RESET  Negative-going reset threshold		vs Free-air temperature	39
V <sub>IT</sub> _			vs Free-air temperature	40
OL(RESET)	RESET output current		vs Input voltage	41
t <sub>d</sub>	Reset time delay		vs Free-air temperature	42
t <sub>d</sub>	Distribution for reset delay			43

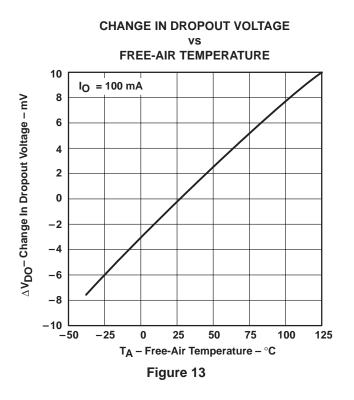


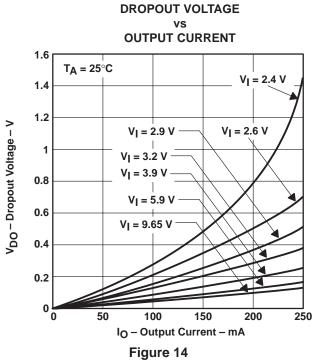


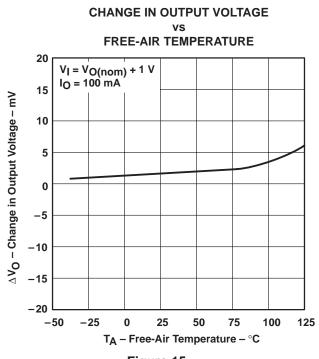




**TPS7301** 









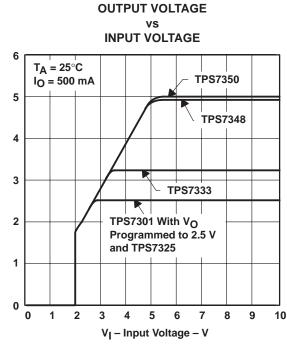
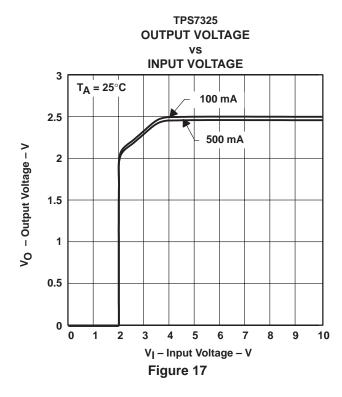
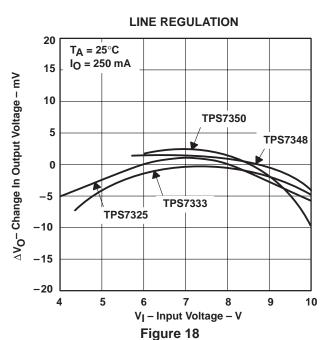
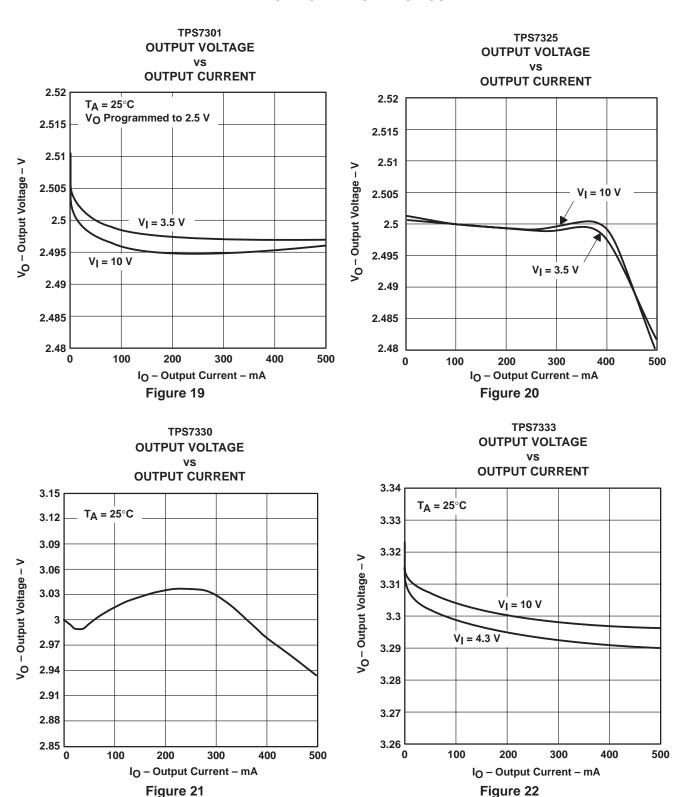


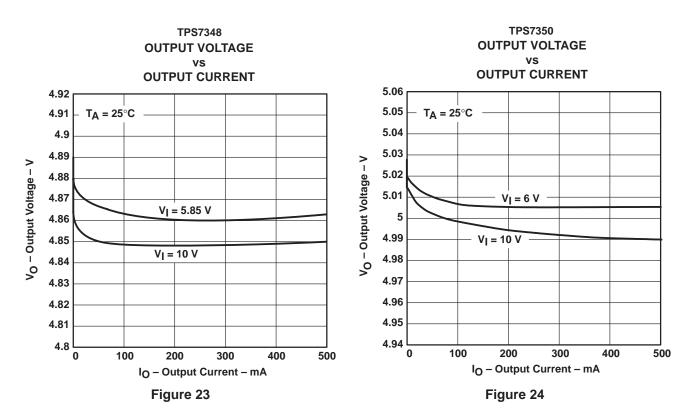
Figure 15













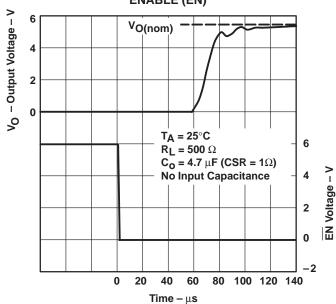


Figure 25



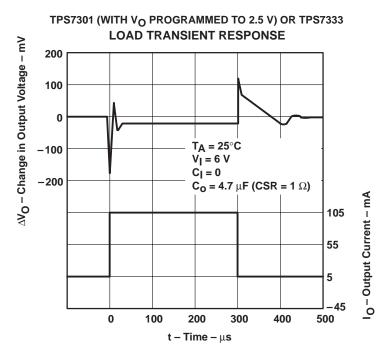


Figure 26

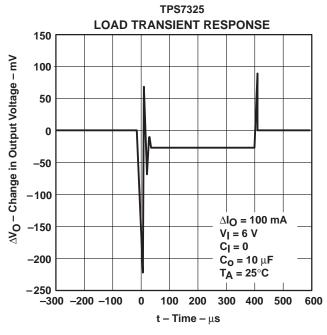


Figure 27

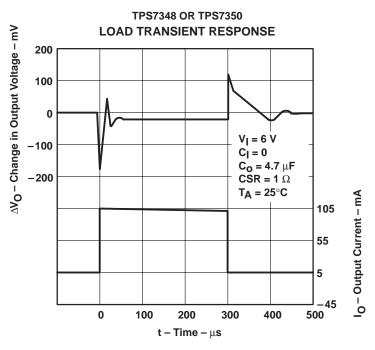


Figure 28

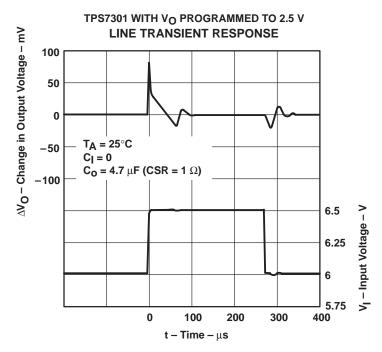


Figure 29

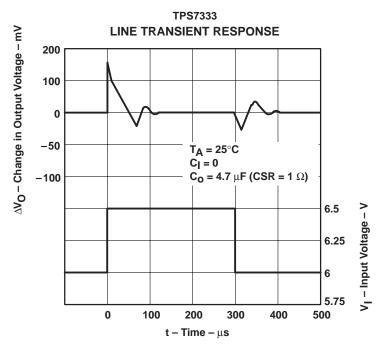


Figure 30

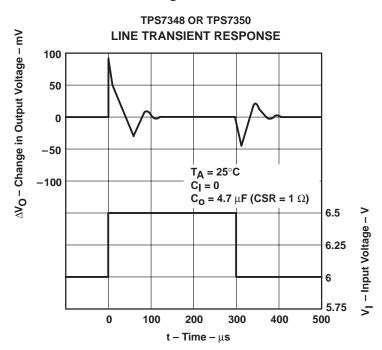


Figure 31

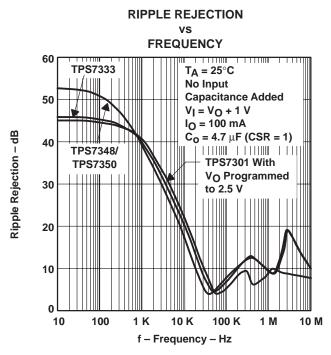
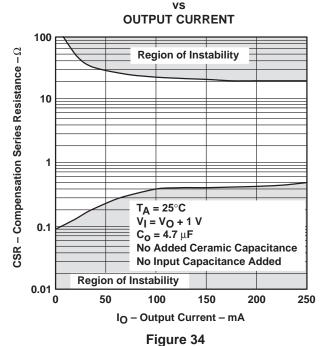


Figure 32

## TYPICAL REGIONS OF STABILITY COMPENSATION SERIES RESISTANCE (CSR)†



**OUTPUT SPECTRAL-NOISE DENSITY** 

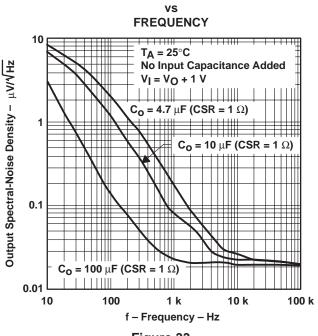


Figure 33

## TYPICAL REGIONS OF STABILITY COMPENSATION SERIES RESISTANCE (CSR) $^{\dagger}$

### ADDED CERAMIC CAPACITANCE

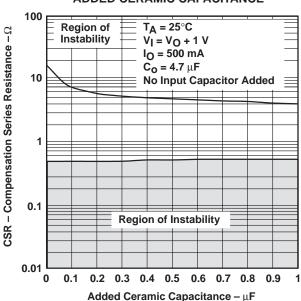


Figure 35



### TYPICAL CHARACTERISTICS

### TYPICAL REGIONS OF STABILITY COMPENSATION SERIES RESISTANCE (CSR)†

### vs **OUTPUT CURRENT** 100 Region of Instability CSR – Compensation Series Resistance – $\Omega$ 10 $T_A = 25^{\circ}C$ VI = VO + 1 V $C_0 = 10 \mu F$ No Added Ceramic Capacitance No Input Capacitor Added 0.1 Region of Instability 0.01 50 0 100 150 200 250 IO - Output Current - mA

Figure 36

## TYPICAL REGIONS OF STABILITY **COMPENSATION SERIES RESISTANCE (CSR)**† ADDED CERAMIC CAPACITANCE

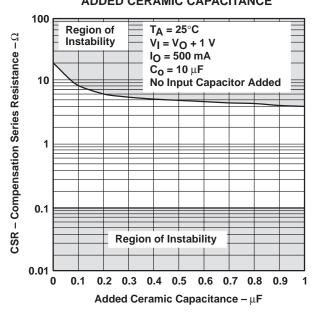
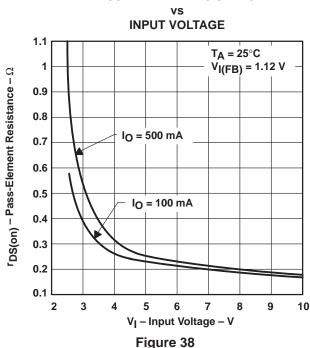


Figure 37

### **PASS-ELEMENT RESISTANCE**



## MINIMUM INPUT VOLTAGE FOR VALID RESET

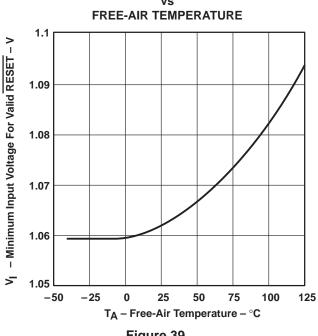
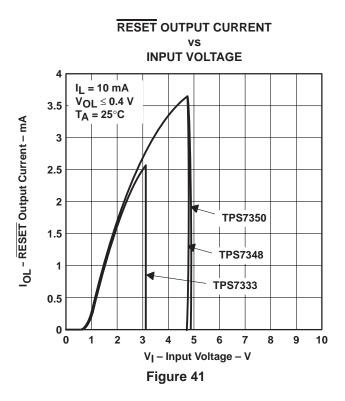
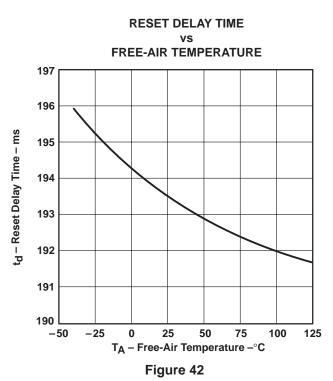
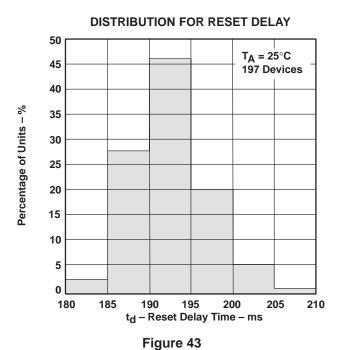


Figure 39

## **NEGATIVE-GOING RESET THRESHOLD** FREE-AIR TEMPERATURE 15 V<sub>IT</sub> - Negative-Going Reset Threshold - mV 10 5 0 -5 -10-25 25 50 75 100 125 -50 $T_A$ – Free-Air Temperature – $^{\circ}C$ Figure 40







TEXAS INSTRUMENTS

SLVS124F – JUNE 1995 – REVISED JANUARY 1999

#### THERMAL INFORMATION

In response to system-miniaturization trends, integrated circuits are being offered in low-profile and fine-pitch surface-mount packages. Implementation of many of today's high-performance devices in these packages requires special attention to power dissipation. Many system-dependent issues such as thermal coupling, airflow, added heat sinks and convection surfaces, and the presence of other heat-generating components affect the power-dissipation limits of a given component.

Three basic approaches for enhancing thermal performance are illustrated in this discussion:

- Improving the power-dissipation capability of the PWB design
- Improving the thermal coupling of the component to the PWB
- Introducing airflow in the system

Figure 44 is an example of a thermally enhanced PWB layout for the 20-lead TSSOP package. This layout involves adding copper on the PWB to conduct heat away from the device. The  $R_{\theta JA}$  (thermal resistance, junction-to-ambient) for this component/board system is illustrated in Figure 45. The family of curves illustrates the effect of increasing the size of the copper-heat-sink surface area. The PWB is a standard FR4 board (L × W × H = 3.2 inch × 3.2 inch × 0.062 inch); the board traces and heat sink area are 1-oz (per square foot) copper.

Figure 46 shows the thermal resistance for the same system with the addition of a thermally-conductive compound between the body of the TSSOP package and the PWB copper routed directly beneath the device. The thermal conductivity for the compound used in this analysis is  $0.815 \text{ W/m} \times ^{\circ}\text{C}$ .

Using these figures to determine the system  $R_{\theta JA}$  allows the maximum power-dissipation limit to be calculated with the equation:

$$P_{D(max)} = \frac{T_{J(max)} - T_{A}}{R_{\theta JA(system)}}$$

Where

T<sub>J(max)</sub> is the maximum allowable junction temperature; 150°C absolute maximum and 125°C maximum recommended operating temperature for specified operation.

This limit should then be applied to the internal power dissipated by the TPS73xx regulator. The equation for calculating total internal power dissipation of the TPS73xx is:

$$P_{D(total)} = (V_I - V_O) \times I_O + V_I \times I_O$$

Because the quiescent current of the TPS73xx family is very low, the second term is negligible, further simplifying the equation to:

$$P_{D(total)} = (V_I - V_O) \times I_O$$

For a 20-lead TSSOP/FR4 board system with thermally conductive compound between the board and the device body, where  $T_A = 55$ °C, airflow = 100 ft/min, and copper heat sink area = 1 cm<sup>2</sup>, the maximum power-dissipation limit can be calculated. As indicated in Figure 46, the system  $R_{\theta JA}$  is 94°C/W; therefore, the maximum power-dissipation limit is:

$$P_{D(max)} = \frac{T_{J(max)} - T_{A}}{R_{\theta JA(system)}} = \frac{125^{\circ}C - 55^{\circ}C}{94^{\circ}C/W} = 745 \text{ mW}$$

If the system implements a TPS7348 regulator where  $V_I = 6 \text{ V}$  and  $I_O = 150 \text{ mA}$ , the internal power dissipation is:

$$P_{D(total)} = (V_1 - V_0) \times I_0 = (6 - 4.85) \times 0.150 = 173 \text{ mW}$$



SLVS124F - JUNE 1995 - REVISED JANUARY 1999

### THERMAL INFORMATION

Comparing  $P_{D(total)}$  with  $P_{D(max)}$  reveals that the power dissipation in this example does not exceed the maximum limit. When it does, one of two corrective actions can be taken. The power-dissipation limit can be raised by increasing either the airflow or the heat-sink area. Alternatively, the internal power dissipation of the regulator can be lowered by reducing either the input voltage or the load current. In either case, the above calculations should be repeated with the new system parameters.

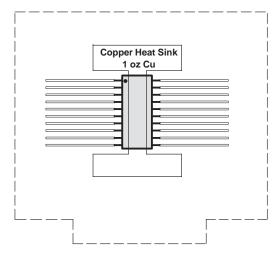
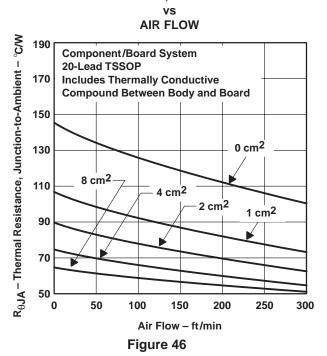


Figure 44. Thermally Enhanced PWB Layout (not to scale) for the 20-Pin TSSOP

### THERMAL RESISTANCE, JUNCTION-TO-AMBIENT

#### vs **AIR FLOW** $R_{\theta JA}$ – Thermal Resistance, Junction-to-Ambient – $^{\circ}$ C/W 190 Component/Board System 20-Lead TSSOP 170 $0 \text{ cm}^2$ 1 cm<sup>2</sup> 150 2 cm<sup>2</sup> 130 110 90 4 cm<sup>2</sup> 8 cm<sup>2</sup> 70 50 50 100 150 200 0 250 300 Air Flow - ft/min Figure 45

#### THERMAL RESISTANCE, JUNCTION-TO-AMBIENT



SLVS124F – JUNE 1995 – REVISED JANUARY 1999

### APPLICATION INFORMATION

The TPS73xx series of low-dropout (LDO) regulators overcome many of the shortcomings of earlier generation LDOs, while adding features such as a power-saving shutdown mode and a supply-voltage supervisor. The TPS73xx family includes five fixed-output voltage regulators: the TPS7325 (2.5 V), TPS7330 (3 V), TPS7333 (3.3 V), the TPS7348 (4.85 V), and the TPS7350 (5 V). The family also offers an adjustable device, the TPS7301 (adjustable from 1.2 V to 9.75 V).

### device operation

The TPS73xx, unlike many other LDOs, features very low quiescent currents that remain virtually constant even with varying loads. Conventional LDO regulators use a pnp-pass element, the base current of which is directly proportional to the load current through the regulator ( $I_B = I_C/\beta$ ). Close examination of the data sheets reveals that such devices are typically specified under near no-load conditions; actual operating currents are much higher as evidenced by typical quiescent current versus load current curves (see Figure 7). The TPS73xx uses a PMOS transistor to pass current; because the gate of the PMOS element is voltage driven, operating currents are low and invariable over the full load range. The TPS73xx specifications reflect actual performance under load.

Another pitfall associated with the pnp-pass element is its tendency to saturate when the device goes into dropout. The resulting drop in  $\beta$  forces an increase in  $I_B$  to maintain the load. During power-up, this translates to large start-up currents. Systems with limited supply current may fail to start up. In battery-powered systems, it means rapid battery discharge when the voltage decays below the minimum required for regulation. The TPS73xx quiescent current remains low even when the regulator drops out, thus eliminating both problems.

Included in the TPS73xx family is a 4.85-V regulator, the TPS7348. Designed specifically for 5-V cellular systems, its 4.85-V output, regulated to within  $\pm$  2%, allows for operation within the low-end limit of 5-V systems specified to  $\pm$  5% tolerance; therefore, maximum regulated operating lifetime is obtained from a battery pack before the device drops out, adding crucial talk minutes between charges.

The TPS73xx family also features a shutdown mode that places the output in the high-impedance state (essentially equal to the feedback-divider resistance) and reduces quiescent current to under  $0.5 \,\mu\text{A}$ . When the shutdown feature is not used,  $\overline{\text{EN}}$  should be tied to ground. Response to an enable transition is quick; regulated output voltage is reestablished in typically 120  $\mu\text{s}$ .

### minimum load requirements

The TPS73xx family is stable even at zero load; no minimum load is required for operation.

### **SENSE** connection

The SENSE terminal of fixed-output devices must be connected to the regulator output for proper functioning of the regulator. Normally, this connection should be as short as possible; however, the connection can be made near a critical circuit (remote sense) to improve performance at that point. Internally, SENSE connects to a high-impedance wide-bandwidth amplifier through a resistor-divider network, and noise pickup feeds through to the regulator output. It is essential to route the SENSE connection in such a way as to minimize/avoid noise pickup. Adding an RC network between SENSE and OUT to filter noise is not recommended because it can cause the regulator to oscillate.

### external capacitor requirements

An input capacitor is not required; however, a ceramic bypass capacitor (0.047 pF to 0.1  $\mu$ F) improves load transient response and noise rejection when the TPS73xx is located more than a few inches from the power supply. A higher-capacitance electrolytic capacitor may be necessary if large (hundreds of milliamps) load transients with fast rise times are anticipated.



### **APPLICATION INFORMATION**

### external capacitor requirements (continued)

As with most LDO regulators, the TPS73xx family requires an output capacitor for stability. A low-ESR 10- $\mu$ F solid-tantalum capacitor connected from the regulator output to ground is sufficient to ensure stability over the full load range (see Figure 42). Adding high-frequency ceramic or film capacitors (such as power-supply bypass capacitors for digital or analog ICs) can cause the regulator to become unstable unless the ESR of the tantalum capacitor is less than 1.2  $\Omega$  over temperature. Capacitors with published ESR specifications such as the AVX TPSD106M035R0300 and the Sprague 593D106X0035D2W work well because the maximum ESR at 25°C is 300 m $\Omega$  (typically, the ESR in solid-tantalum capacitors increases by a factor of 2 or less when the temperature drops from 25°C to -40°C). Where component height and/or mounting area is a problem, physically smaller, 10- $\mu$ F devices can be screened for ESR. Figures 29 through 32 show the stable regions of operation using different values of output capacitance with various values of ceramic load capacitance.

In applications with little or no high-frequency bypass capacitance (< 0.2  $\mu$ F), the output capacitance can be reduced to 4.7  $\mu$ F, provided ESR is maintained between 0.7 and 2.5  $\Omega$ . Because capacitor minimum ESR is seldom if ever specified, it may be necessary to add a 0.5- $\Omega$  to 1- $\Omega$  resistor in series with the capacitor and limit ESR to 1.5  $\Omega$  maximum. As shown in the CSR graphs (Figures 29 through 32), minimum ESR is not a problem when using 10- $\mu$ F or larger output capacitors.

Below is a partial listing of surface-mount capacitors usable with the TPS73xx family. This information, along with the CSR graphs, is included to assist in selection of suitable capacitance for the user's application. When necessary to achieve low height requirements along with high output current and/or high ceramic load capacitance, several higher ESR capacitors can be used in parallel to meet the guidelines above.

All load and temperature conditions with up to 1  $\mu F$  of added ceramic load capacitance:

PART NO.	MFR.	VALUE	MAX ESR†	SIZE $(H \times L \times W)^{\dagger}$
T421C226M010AS	Kemet	$22~\mu\text{F},~10~\text{V}$	0.5	$2.8\times 6\times 3.2$
593D156X0025D2W	Sprague	15 μF, 25 V	0.3	$2.8\times7.3\times4.3$
593D106X0035D2W	Sprague	10 μF, 35 V	0.3	$2.8\times7.3\times4.3$
TPSD106M035R0300	AVX	10 μF, 35 V	0.3	$2.8 \times 7.3 \times 4.3$

Load < 200 mA, ceramic load capacitance < 0.2 μF, full temperature range:

PART NO.	MFR.	VALUE	MAX ESR†	SIZE $(H \times L \times W)^{\dagger}$
592D156X0020R2T	Sprague	15 μF, 20 V	1.1	$1.2\times7.2\times6$
595D156X0025C2T	Sprague	15 μF, 25 V	1	$2.5\times7.1\times3.2$
595D106X0025C2T	Sprague	10 $\mu F$ , 25 $V$	1.2	$2.5\times7.1\times3.2$
293D226X0016D2W	Sprague	22 μF, 16 V	1.1	$2.8 \times 7.3 \times 4.3$

Load < 100 mA, ceramic load capacitance < 0.2  $\mu$ F, full temperature range:

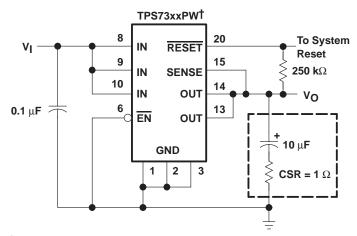
PART NO.	MFR.	VALUE	MAX ESR†	SIZE $(H \times L \times W)^{\dagger}$
195D106X06R3V2T	Sprague	10 $\mu$ F, 6.3 V	1.5	$1.3\times3.5\times2.7$
195D106X0016X2T	Sprague	10 $\mu$ F, 16 V	1.5	$1.3\times7\times2.7$
595D156X0016B2T	Sprague	15 μF, 16 V	1.8	$1.6\times3.8\times2.6$
695D226X0015F2T	Sprague	$22~\mu\text{F},~15~\text{V}$	1.4	$1.8\times6.5\times3.4$
695D156X0020F2T	Sprague	15 $\mu$ F, 20 V	1.5	$1.8\times6.5\times3.4$
695D106X0035G2T	Sprague	10 $\mu$ F, 35 V	1.3	$2.5\times7.6\times2.5$

<sup>†</sup> Size is in mm. ESR is maximum resistance at 100 kHz and T<sub>A</sub> = 25°C. Listings are sorted by height.



### **APPLICATION INFORMATION**

### external capacitor requirements (continued)



† TPS7333, TPS7348, TPS7350 (fixed-voltage options)

Figure 47. Typical Application Circuit

### programming the TPS7301 adjustable LDO regulator

Programming the adjustable regulators is accomplished using an external resistor divider as shown in Figure 43. The equation governing the output voltage is:

$$V_O = V_{ref} \times \left(1 + \frac{R1}{R2}\right)$$

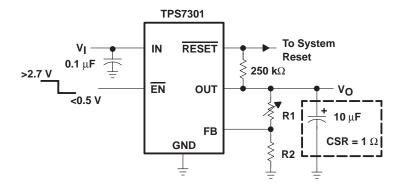
Where

V<sub>ref</sub> = reference voltage, 1.182 V typ

### **APPLICATION INFORMATION**

Resistors R1 and R2 should be chosen for approximately  $7-\mu A$  divider current. A recommended value for R2 is 169 k $\Omega$  with R1 adjusted for the desired output voltage. Smaller resistors can be used, but offer no inherent advantage and consume more power. Larger values of R1 and R2 should be avoided as leakage currents at FB will introduce an error. Solving for R1 yields a more useful equation for choosing the appropriate resistance:

$$R1 = \left(\frac{V_O}{V_{ref}} - 1\right) \times R2$$



## OUTPUT VOLTAGE

I KOOKAMIMINO OOIDE						
OUTPUT VOLTAGE	R1	R2	UNIT			
2.5 V	191	169	kΩ			
3.3 V	309	169	kΩ			
3.6 V	348	169	kΩ			
4 V	402	169	kΩ			
5 V	549	169	kΩ			
6.4 V	750	169	kΩ			

Figure 48. TPS7301 Adjustable LDO Regulator Programming

### undervoltage supervisor function

The RESET output of the TPS73xx initiates a reset in microcomputer and microprocessor systems in the event of an undervoltage condition. An internal comparator in the TPS73xx monitors the output voltage of the regulator to detect the undervoltage condition. When that occurs, the RESET output transistor turns on taking the RESET signal low.

On power up, the output voltage tracks the input voltage. The  $\overline{\text{RESET}}$  output becomes active (low) as  $V_I$  approaches the minimum required for a valid  $\overline{\text{RESET}}$  signal (specified at 1.5 V for 25°C and 1.9 V over full recommended operating temperature range). When the output voltage reaches the appropriate positive-going input threshold ( $V_{IT+}$ ), a 200-ms (typical) timeout period begins during which the  $\overline{\text{RESET}}$  output remains low. Once the timeout has expired, the  $\overline{\text{RESET}}$  output becomes inactive. Since the  $\overline{\text{RESET}}$  output is an open-drain NMOS, a pullup resistor should be used to ensure that a logic-high signal is indicated.

The supply-voltage-supervisor function is also activated during power-down. As the input voltage decays and after the dropout voltage is reached, the output voltage tracks linearly with the decaying input voltage. When the output voltage drops below the specified negative-going input threshold ( $V_{IT-}$  — see electrical characteristics tables), the  $\overline{RESET}$  output becomes active (low). It is important to note that if the input voltage decays below the minimum required for a valid  $\overline{RESET}$ , the  $\overline{RESET}$  is undefined.

Since the circuit is monitoring the regulator output voltage, the  $\overline{RESET}$  output can also be triggered by disabling the regulator or by any fault condition that causes the output to drop below  $V_{IT}$ . Examples of fault conditions include a short circuit on the output and a low input voltage. Once the output voltage is reestablished, either by reenabling the regulator or removing the fault condition, then the internal timer is initiated, which holds the  $\overline{RESET}$  signal active during the 200-ms (typical) timeout period.



SLVS124F - JUNE 1995 - REVISED JANUARY 1999

### **APPLICATION INFORMATION**

### undervoltage supervisor function (continued)

Transient loads or line pulses can also cause a reset to occur if proper care is not taken in selecting the input and output capacitors. Load transients that are faster than 5  $\mu$ s can cause a reset if high-ESR output capacitors (greater than approximately 7  $\Omega$ ) are used. A 1- $\mu$ s transient causes a reset when using an output capacitor with greater than 3.5  $\Omega$  of ESR. Note that the output-voltage spike during the transient can drop well below the reset threshold and still not trip if the transient duration is short. A 1- $\mu$ s transient must drop at least 500 mV below the threshold before tripping the reset circuit. A 2- $\mu$ s transient trips RESET at just 400 mV below the threshold. Lower-ESR output capacitors help by reducing the drop in output voltage during a transient and should be used when fast transients are expected.

**NOTE:**  $V_{IT+} = V_{IT-} + Hysteresis$ 

### output noise

The TPS73xx has very low output noise, with a spectral noise density < 2  $\mu$ V/ $\sqrt{Hz}$ . This is important when noise-susceptible systems, such as audio amplifiers, are powered by the regulator.

### regulator protection

The TPS73xx PMOS-pass transistor has a built-in back diode that safely conducts reverse currents when the input voltage drops below the output voltage (e.g., during power down). Current is conducted from the output to the input and is not internally limited. If extended reverse voltage is anticipated, external limiting might be appropriate.

The TPS73xx also features internal current limiting and thermal protection. During normal operation, the TPS73xx limits output current to approximately 1 A. When current limiting engages, the output voltage scales back linearly until the overcurrent condition ends. While current limiting is designed to prevent gross device failure, care should be taken not to exceed the power dissipation ratings of the package. If the temperature of the device exceeds 165°C, thermal-protection circuitry shuts it down. Once the device has cooled, regulator operation resumes.



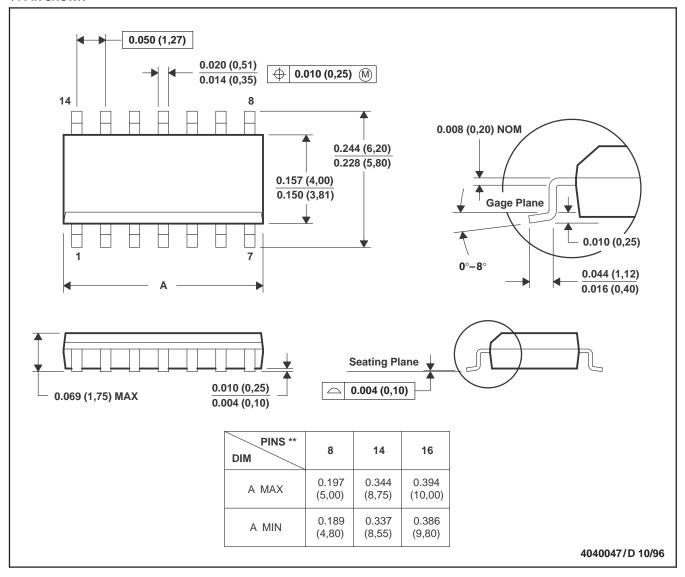
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### **MECHANICAL DATA**

### D (R-PDSO-G\*\*)

### PLASTIC SMALL-OUTLINE PACKAGE

#### 14 PIN SHOWN



NOTES: A. All linear dimensions are in inches (millimeters).

B. This drawing is subject to change without notice.

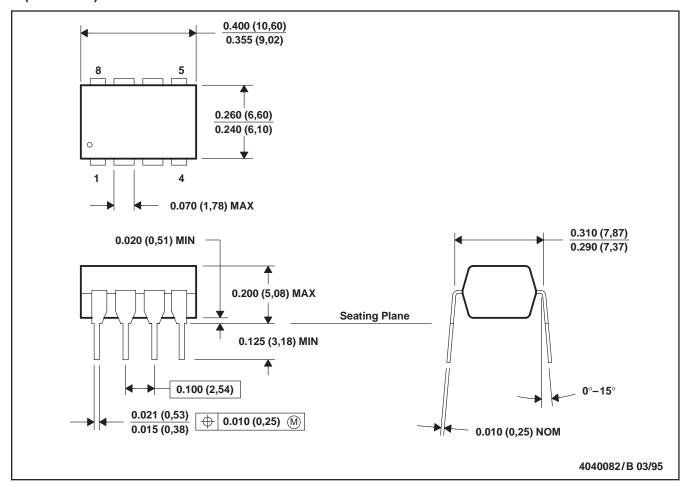
C. Body dimensions do not include mold flash or protrusion, not to exceed 0.006 (0,15).

D. Falls within JEDEC MS-012

### **MECHANICAL DATA**

### P (R-PDIP-T8)

### PLASTIC DUAL-IN-LINE PACKAGE



NOTES: A. All linear dimensions are in inches (millimeters).

- B. This drawing is subject to change without notice.
- C. Falls within JEDEC MS-001

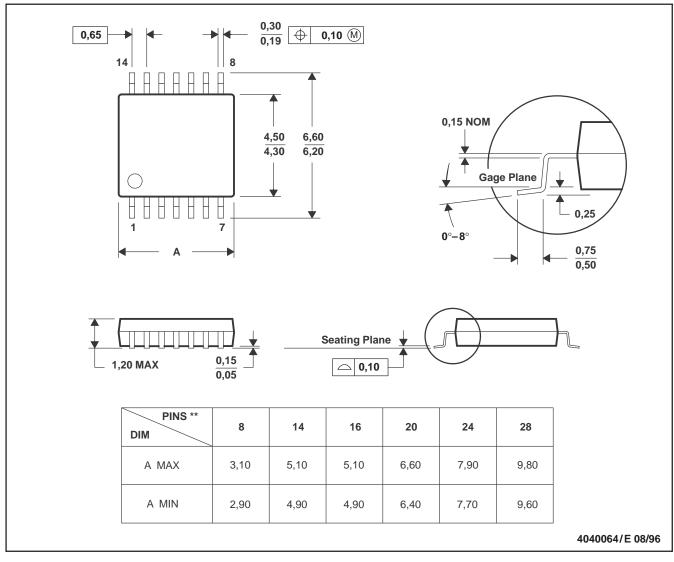
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### **MECHANICAL DATA**

### PW (R-PDSO-G\*\*)

### 14 PIN SHOWN

### PLASTIC SMALL-OUTLINE PACKAGE



NOTES: A. All linear dimensions are in millimeters.

B. This drawing is subject to change without notice.

C. Body dimensions do not include mold flash or protrusion not to exceed 0,15.

D. Falls within JEDEC MO-153



4-Mar-2005





### **PACKAGING INFORMATION**

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins	Package Qty	Eco Plan <sup>(2)</sup>	Lead/Ball Finish	MSL Peak Temp <sup>(3)</sup>
TPS7301QD	ACTIVE	SOIC	D	8	75	Pb-Free (RoHS)	CU NIPDAU	Level-2-260C-1YEAR/ Level-1-220C-UNLIM
TPS7301QDR	ACTIVE	SOIC	D	8	2500	Pb-Free (RoHS)	CU NIPDAU	Level-2-260C-1YEAR/ Level-1-220C-UNLIM
TPS7301QP	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	Level-NC-NC-NC
TPS7301QPW	ACTIVE	TSSOP	PW	20	70	None	CU NIPDAU	Level-1-220C-UNLIM
TPS7301QPWLE	OBSOLETE	TSSOP	PW	20		None	Call TI	Call TI
TPS7301QPWR	ACTIVE	TSSOP	PW	20	2000	None	CU NIPDAU	Level-1-220C-UNLIM
TPS7325QD	ACTIVE	SOIC	D	8	75	Pb-Free (RoHS)	CU NIPDAU	Level-2-260C-1YEAR/ Level-1-220C-UNLIM
TPS7325QDR	ACTIVE	SOIC	D	8	2500	Pb-Free (RoHS)	CU NIPDAU	Level-2-260C-1YEAR/ Level-1-220C-UNLIM
TPS7325QP	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	Level-NC-NC-NC
TPS7325QPW	ACTIVE	TSSOP	PW	20	70	None	CU NIPDAU	Level-1-220C-UNLIM
TPS7325QPWR	ACTIVE	TSSOP	PW	20	2000	None	CU NIPDAU	Level-1-220C-UNLIM
TPS7330QD	ACTIVE	SOIC	D	8	75	Pb-Free (RoHS)	CU NIPDAU	Level-2-260C-1YEAR/ Level-1-220C-UNLIM
TPS7330QDR	ACTIVE	SOIC	D	8	2500	Pb-Free (RoHS)	CU NIPDAU	Level-2-260C-1YEAR/ Level-1-220C-UNLIM
TPS7330QP	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	Level-NC-NC-NC
TPS7330QPWR	ACTIVE	TSSOP	PW	20	2000	None	CU NIPDAU	Level-1-220C-UNLIM
TPS7333QD	ACTIVE	SOIC	D	8	75	None	CU NIPDAU	Level-1-220C-UNLIM
TPS7333QDR	ACTIVE	SOIC	D	8	2500	None	CU NIPDAU	Level-1-220C-UNLIM
TPS7333QP	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	Level-NC-NC-NC
TPS7333QPW	ACTIVE	TSSOP	PW	20	70	None	CU NIPDAU	Level-1-220C-UNLIM
TPS7333QPWLE	OBSOLETE	TSSOP	PW	20		None	Call TI	Call TI
TPS7333QPWR	ACTIVE	TSSOP	PW	20	2000	None	CU NIPDAU	Level-1-220C-UNLIM
TPS7348QD	ACTIVE	SOIC	D	8	75	Pb-Free (RoHS)	CU NIPDAU	Level-2-260C-1YEAR/ Level-1-220C-UNLIM
TPS7348QDR	ACTIVE	SOIC	D	8	2500	Pb-Free (RoHS)	CU NIPDAU	Level-2-260C-1YEAR/ Level-1-220C-UNLIM
TPS7348QP	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	Level-NC-NC-NC
TPS7348QPW	ACTIVE	TSSOP	PW	20	70	None	CU NIPDAU	Level-1-220C-UNLIM
TPS7348QPWLE	OBSOLETE	TSSOP	PW	20		None	Call TI	Call TI
TPS7348QPWR	ACTIVE	TSSOP	PW	20	2000	None	CU NIPDAU	Level-1-220C-UNLIM
TPS7350QD	ACTIVE	SOIC	D	8	75	Pb-Free (RoHS)	CU NIPDAU	Level-2-260C-1YEAR/ Level-1-220C-UNLIM
TPS7350QDR	ACTIVE	SOIC	D	8	2500	Pb-Free (RoHS)	CU NIPDAU	Level-2-260C-1YEAR/ Level-1-220C-UNLIM
TPS7350QP	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	Level-NC-NC-NC
TPS7350QPW	ACTIVE	TSSOP	PW	20	70	None	CU NIPDAU	Level-1-220C-UNLIM



### PACKAGE OPTION ADDENDUM

4-Mar-2005

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing		ckage Qty	Eco Plan <sup>(2)</sup>	Lead/Ball Finish	MSL Peak Temp <sup>(3)</sup>
TPS7350QPWLE	OBSOLETE	TSSOP	PW	20		None	Call TI	Call TI
TPS7350QPWR	ACTIVE	TSSOP	PW	20 2	2000	None	CU NIPDAU	Level-1-220C-UNLIM

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at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

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